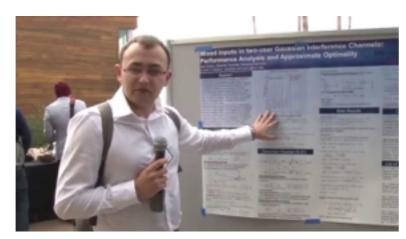
Adding and removing assumptions in network information theory







Natasha Devroye

Alex Dysto

Daniela Tuninetti

Natasha Devroye, Associate Professor

portions based on his Ph.D. slides

Joint work with Alex Dytso, former Ph.D. student, postdoc at Princeton

ELECTRICAL AND COMPUTER ENGINEERING COLLEGE OF ENGINEERING

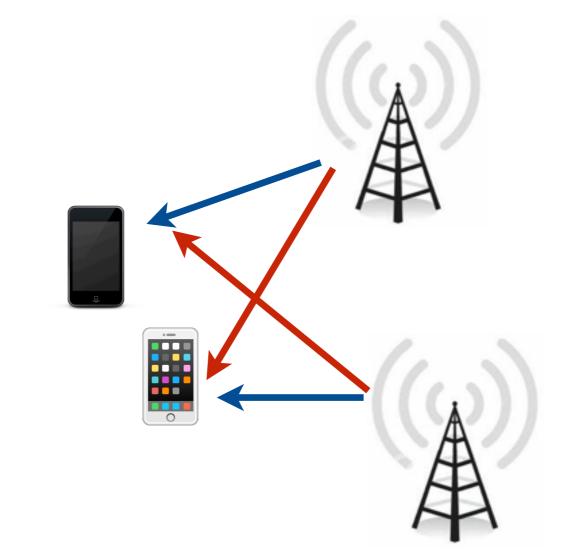
Sara Shahi, current Ph.D. student

Naveed Naimipour, current Ph.D. student

Daniela Tuninetti, Professor

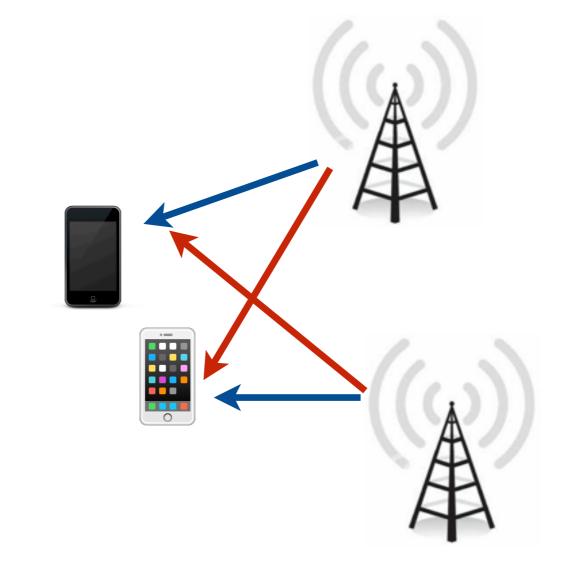
BASIC MODEL FOR INTERFERING WIRELESS CHANNELS:

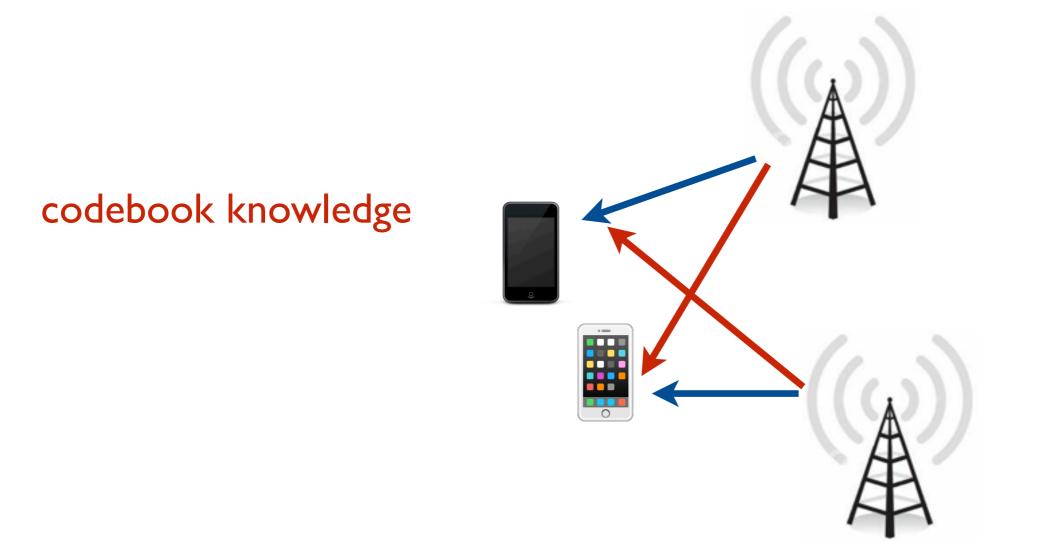
THE INTERFERENCE CHANNEL

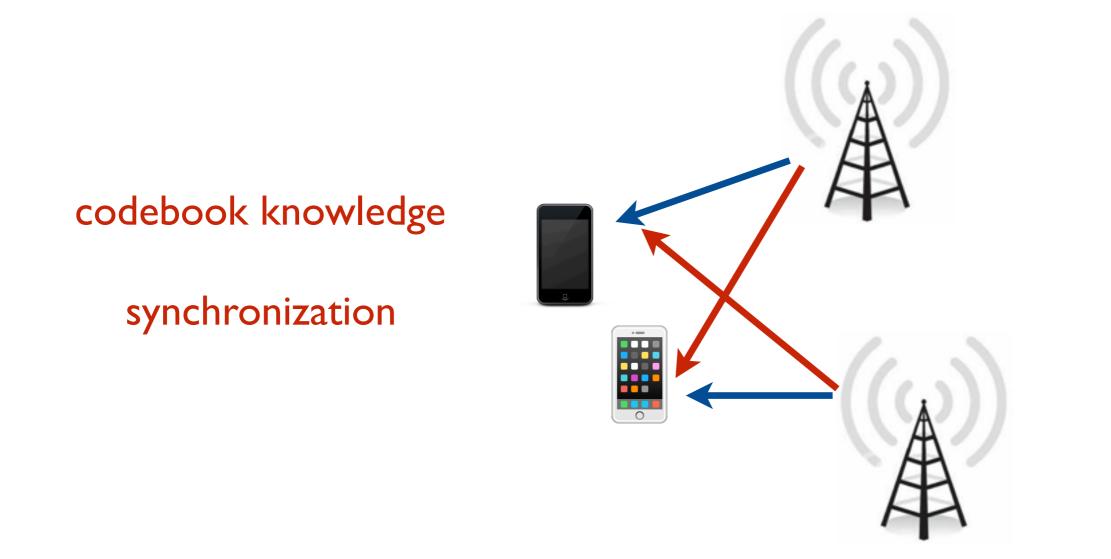


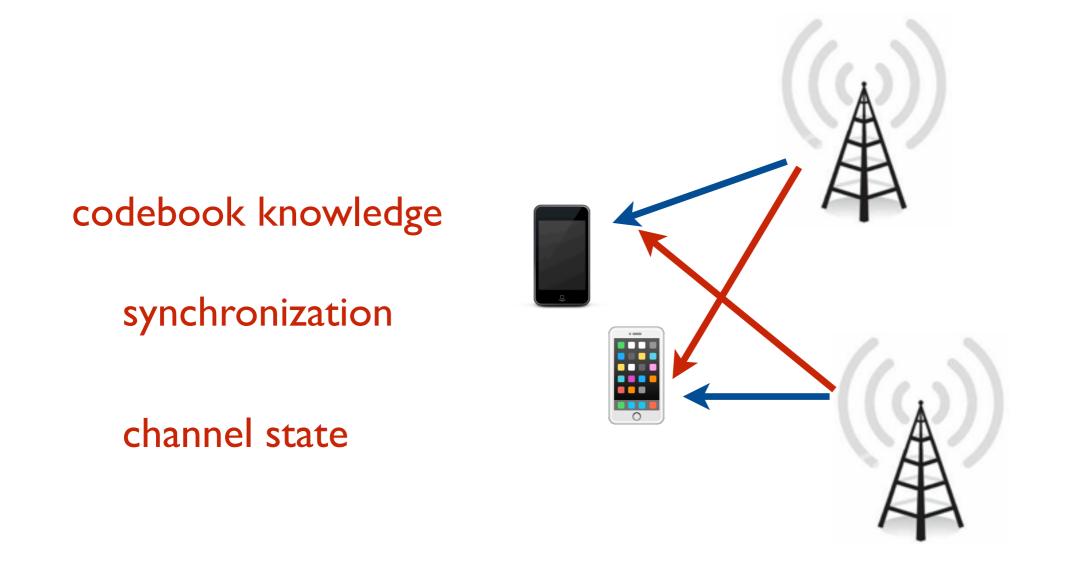
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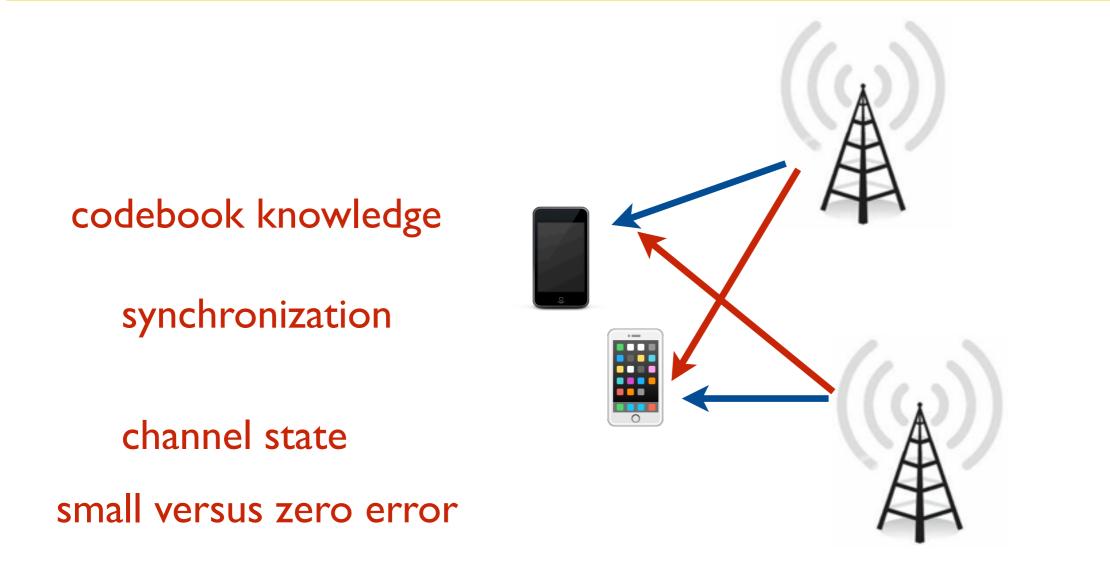
THE INTERFERENCE CHANNEL

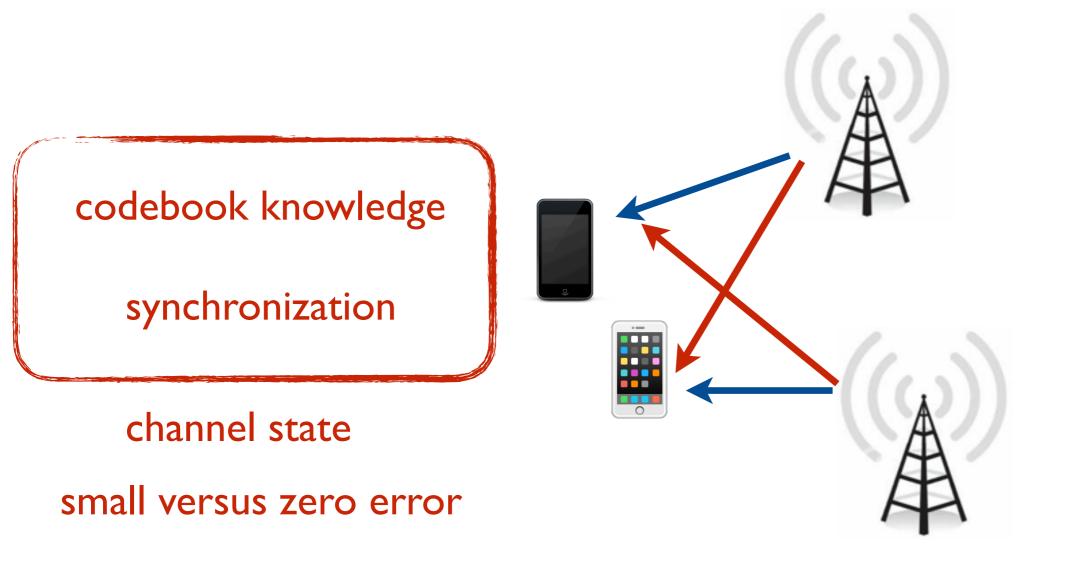




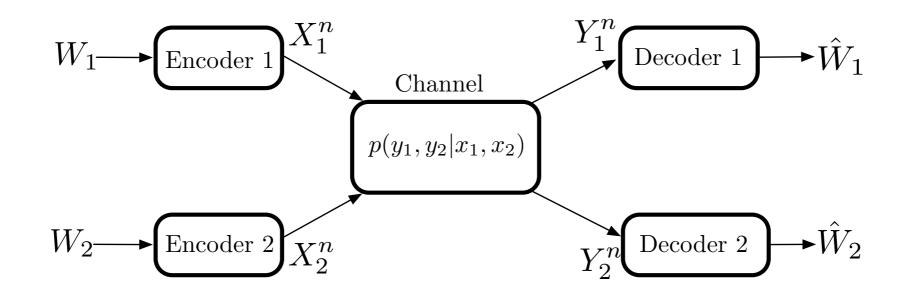






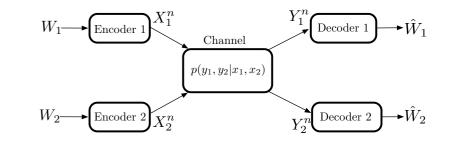


Interference channel



• a discrete memoryless interference channel (DM-IC) $(\mathcal{X}_1 \times \mathcal{X}_2, p(y_1, y_2 | x_1, x_2), \mathcal{Y}_1 \times \mathcal{Y}_2)$ consists of 4 finite sets/alphabets $\mathcal{X}_1, \mathcal{X}_2, \mathcal{Y}_1, \mathcal{Y}_2$ and a collection of conditional pmfs $p(y_1, y_2 | x_1, x_2)$

- sender j = 1, 2 sends an independent message W_i to receiver j
- lower case x is an instance of random variable X in calligraphic alphabet \mathcal{X}



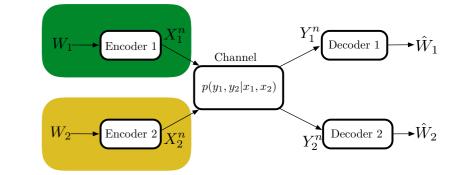
- A $(2^{nR_1}, 2^{nR_2}, n)$ code for the IC consists of:
 - 1. Two message sets $[1:2^{nR_1}]$, and $[1:2^{nR_2}]$
 - 2. Two encoders:

 $w_1 \in [1:2^{nR_1}] \to x_1^n(w_1)$ $w_2 \in [1:2^{nR_2}] \to x_2^n(w_2)$

3. Two decoders:

 $y_1^n \to [1:2^{nR_1}] \cup \text{error}$ $y_2^n \to [1:2^{nR_2}] \cup \text{error}$

- we assume W_1 and W_2 are uniformly distributed on $[1:2^{nR_1}]$ and $[1:2^{nR_2}]$ respectively
- to communicate w_1, w_2 , send $x_1^n(w_1)$ and $x_2^n(w_2)$ over the channel $p(y_1^n, y_2^n | x_1^n, x_2^n)$



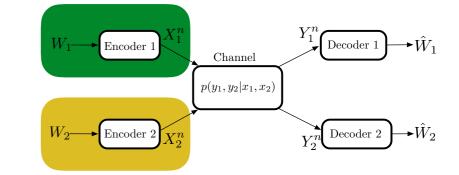
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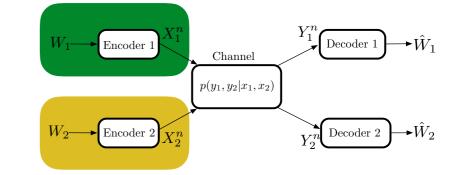
3. Two decoders:

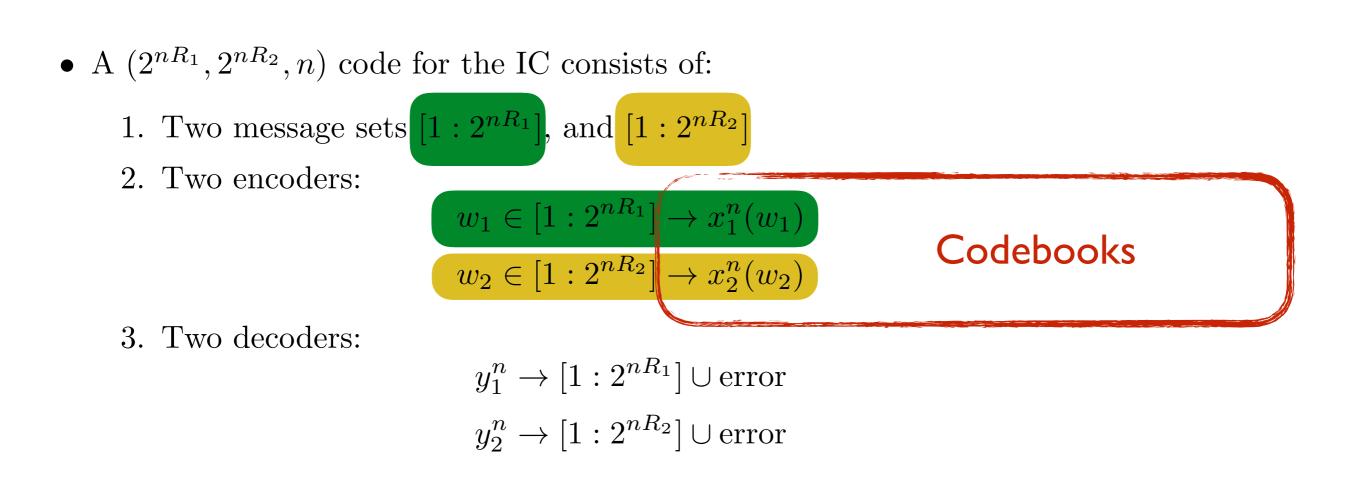
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 $w_1 \in [1:2^{nR_1}] \to x_1^n(w_1)$

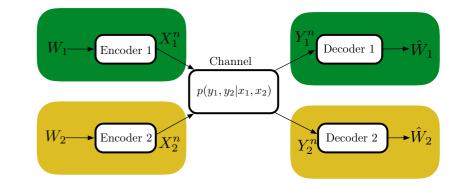
 $w_2 \in [1:2^{nR_2}] \to x_2^n(w_2)$

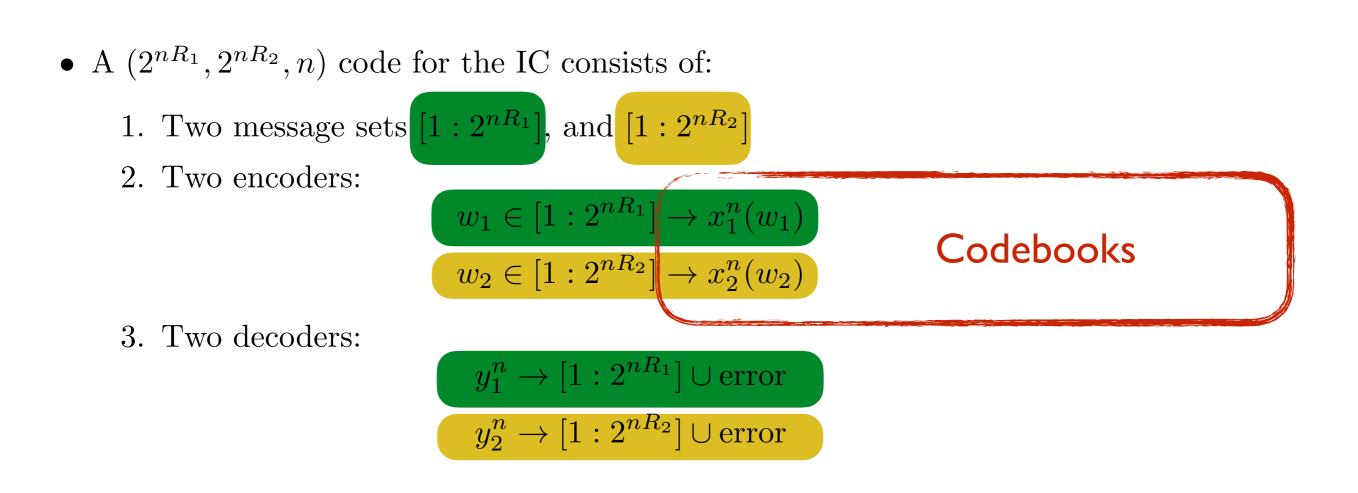
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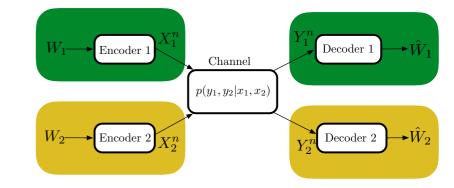


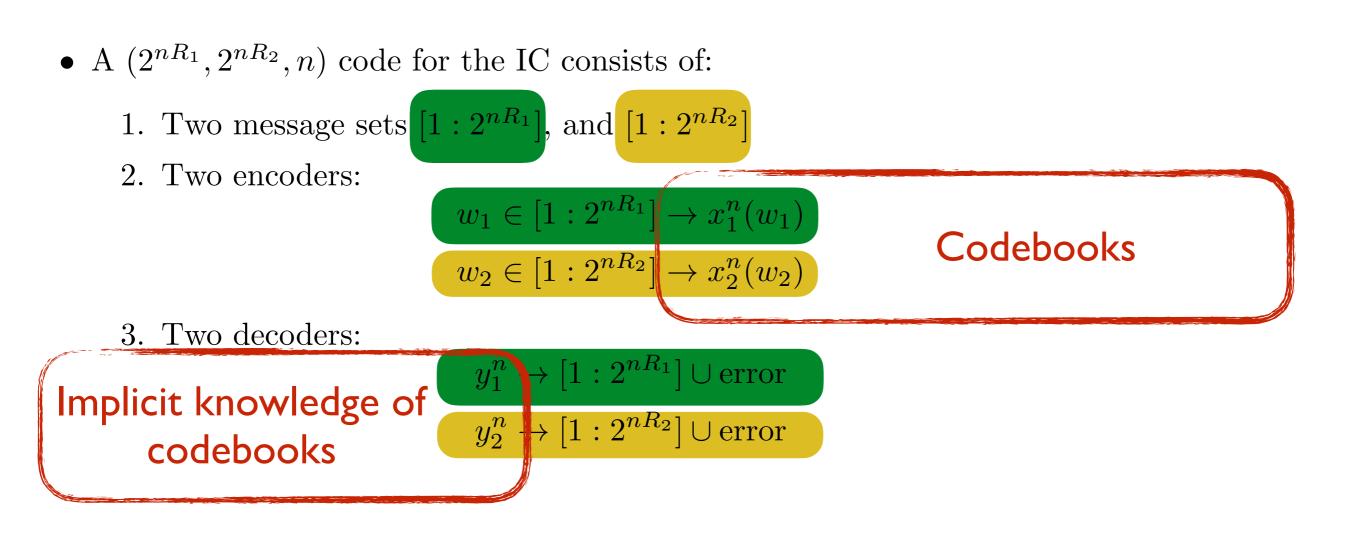
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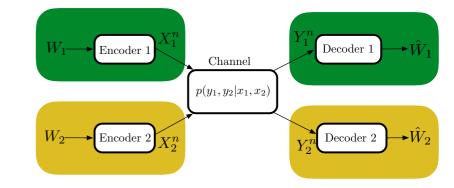


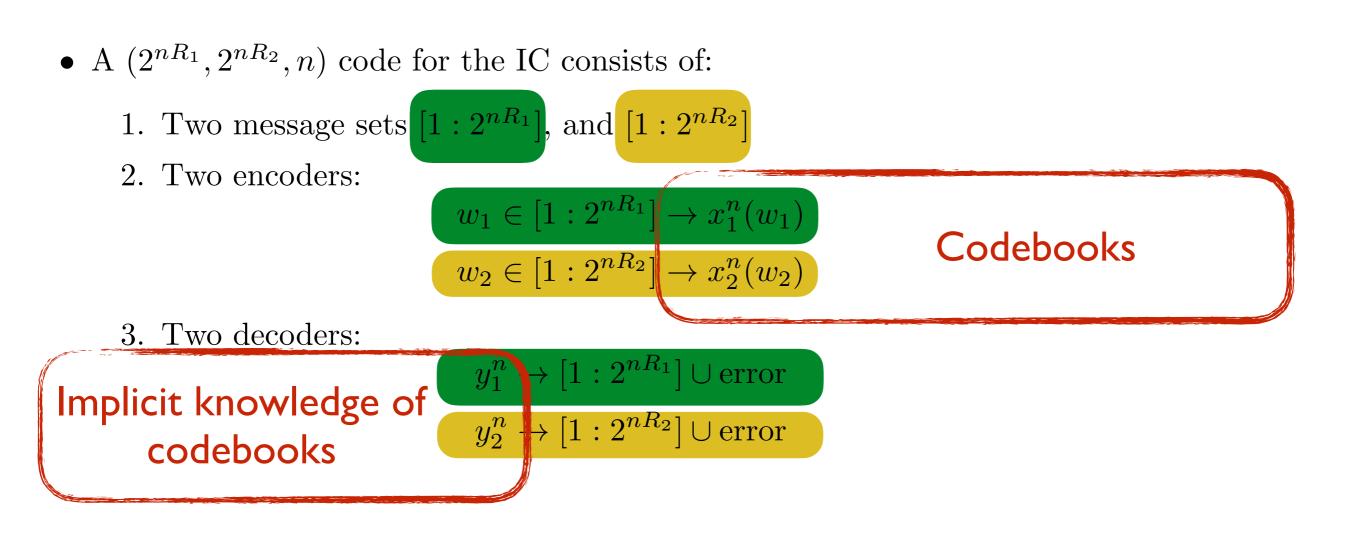
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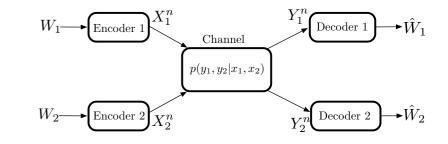


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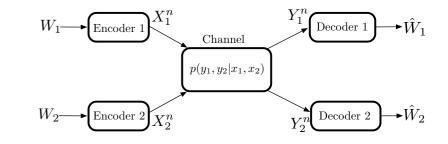


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• average probability of error:

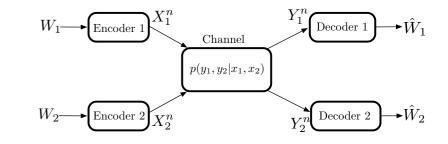
 $P_e^{(n)} := \mathbf{P}\{(\widehat{W}_1, \widehat{W}_2) \neq (W_1, W_2)\}$



• average probability of error:

$$P_e^{(n)} := P\{(\widehat{W}_1, \widehat{W}_2) \neq (W_1, W_2)\}$$

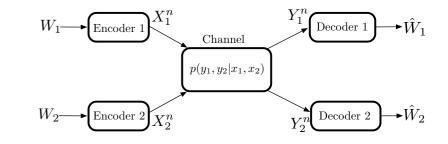
• Rate pair (R_1, R_2) is *achievable* if there exists a sequence of $(2^{nR_1}, 2^{nR_2}, n)$ codes with $P_e^{(n)} \to 0$ as $n \to \infty$



• average probability of error:

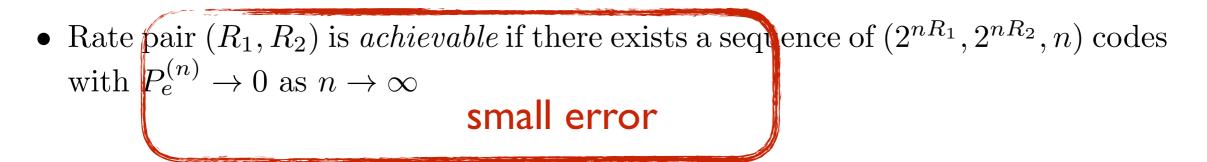
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• Rate pair (R_1, R_2) is *achievable* if there exists a sequence of $(2^{nR_1}, 2^{nR_2}, n)$ codes with $P_e^{(n)} \to 0$ as $n \to \infty$ small error

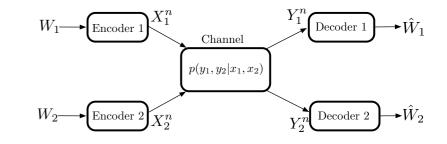


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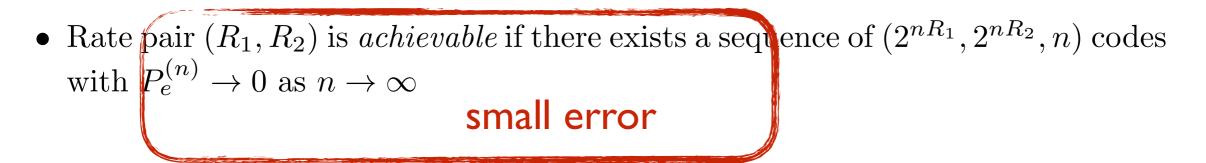


• The capacity region of the DM-IC is the closure of the set of achievable rate pairs (R_1, R_2)



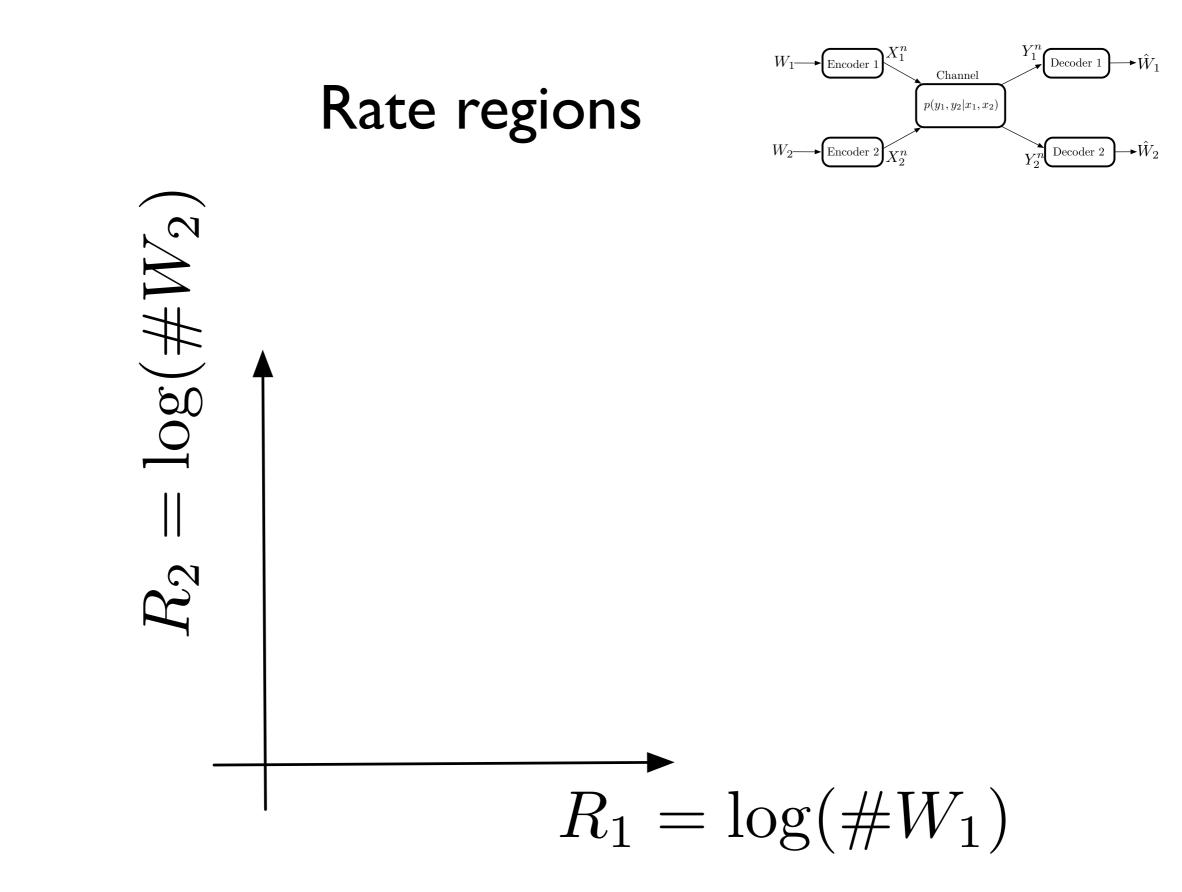
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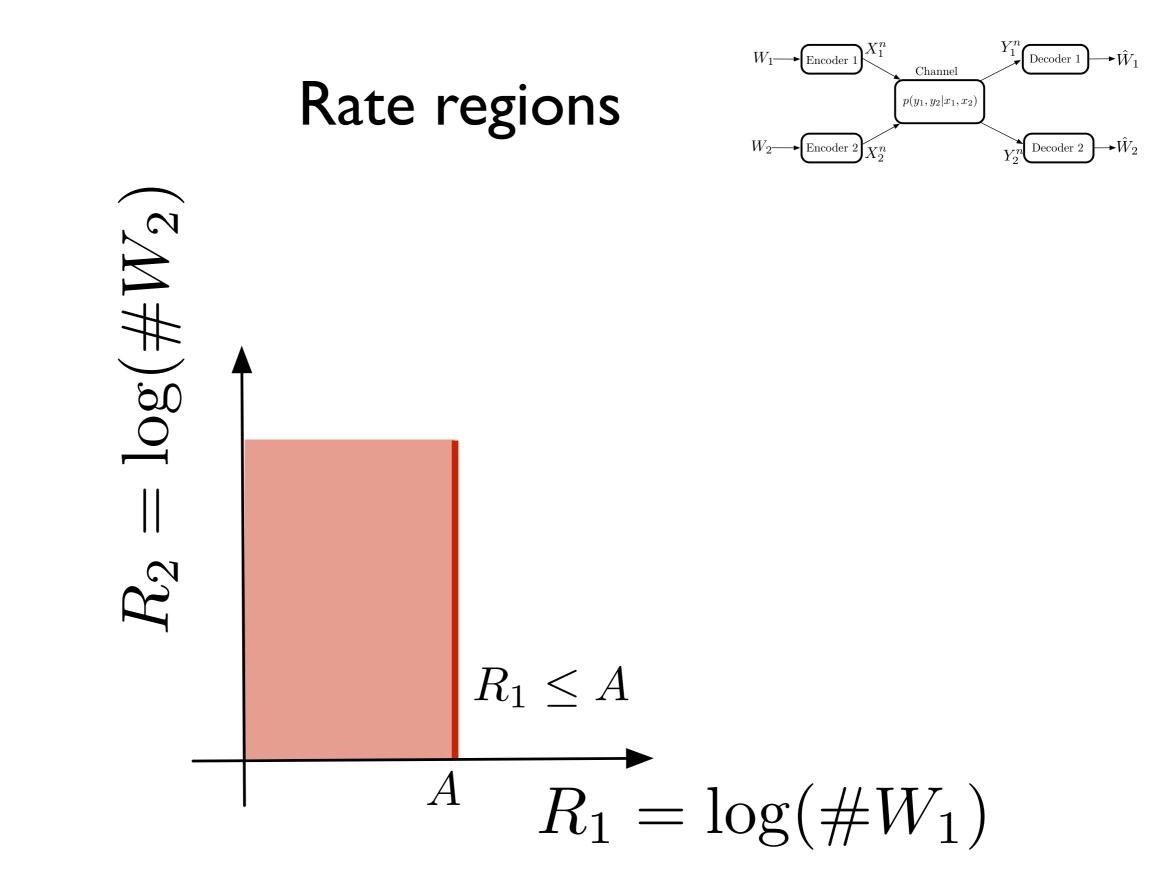
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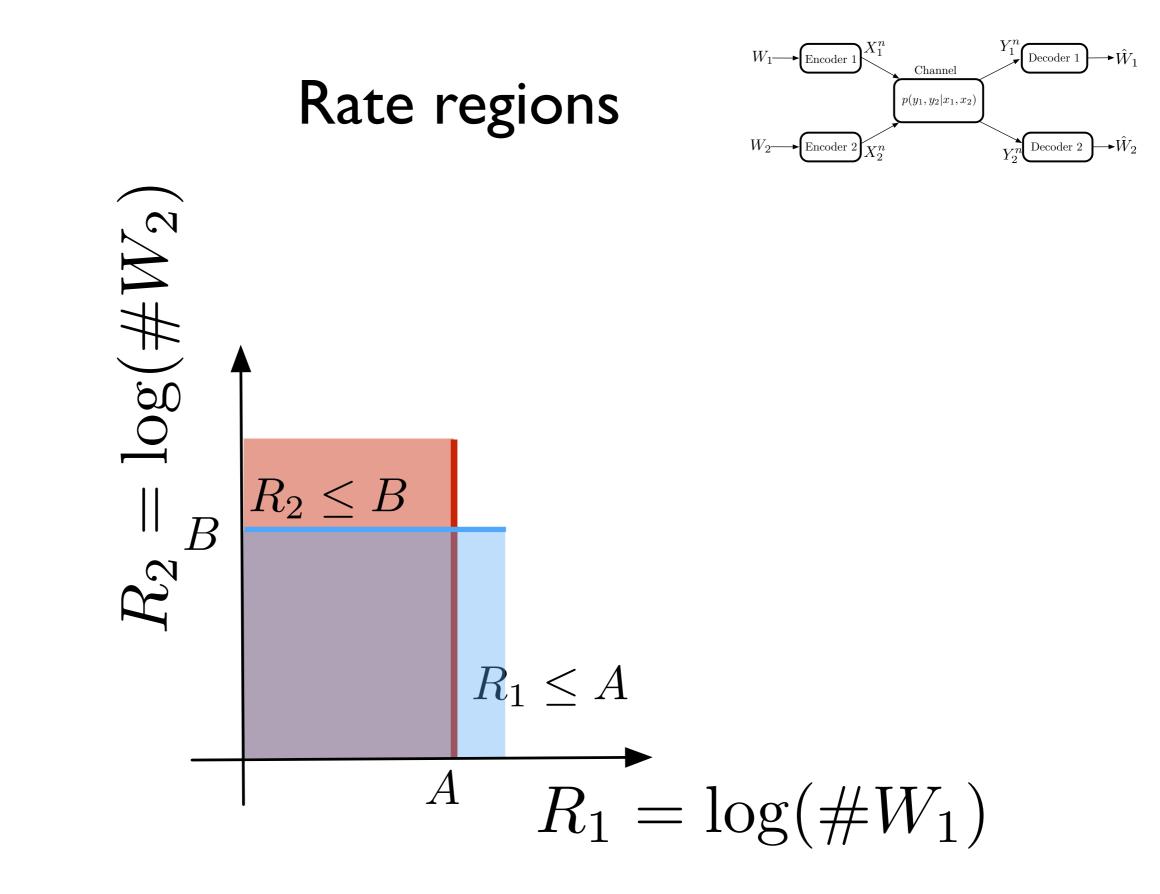


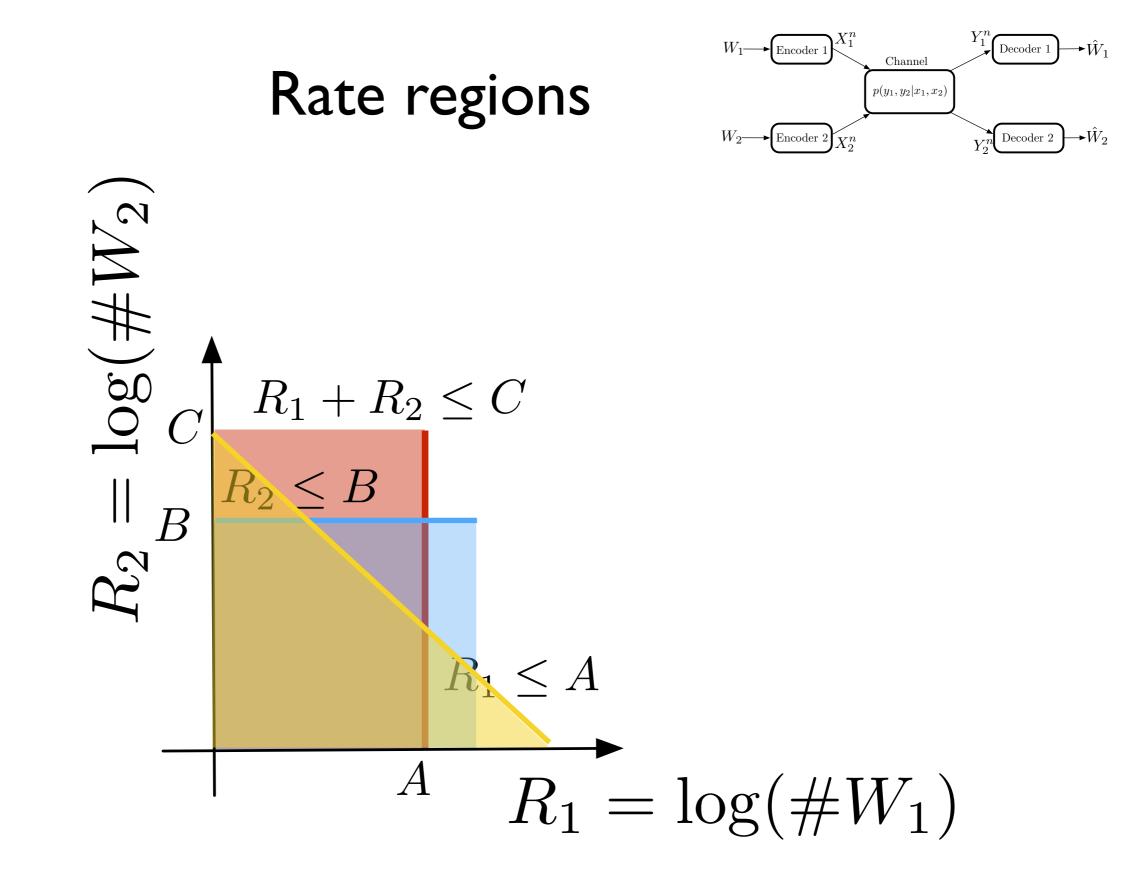
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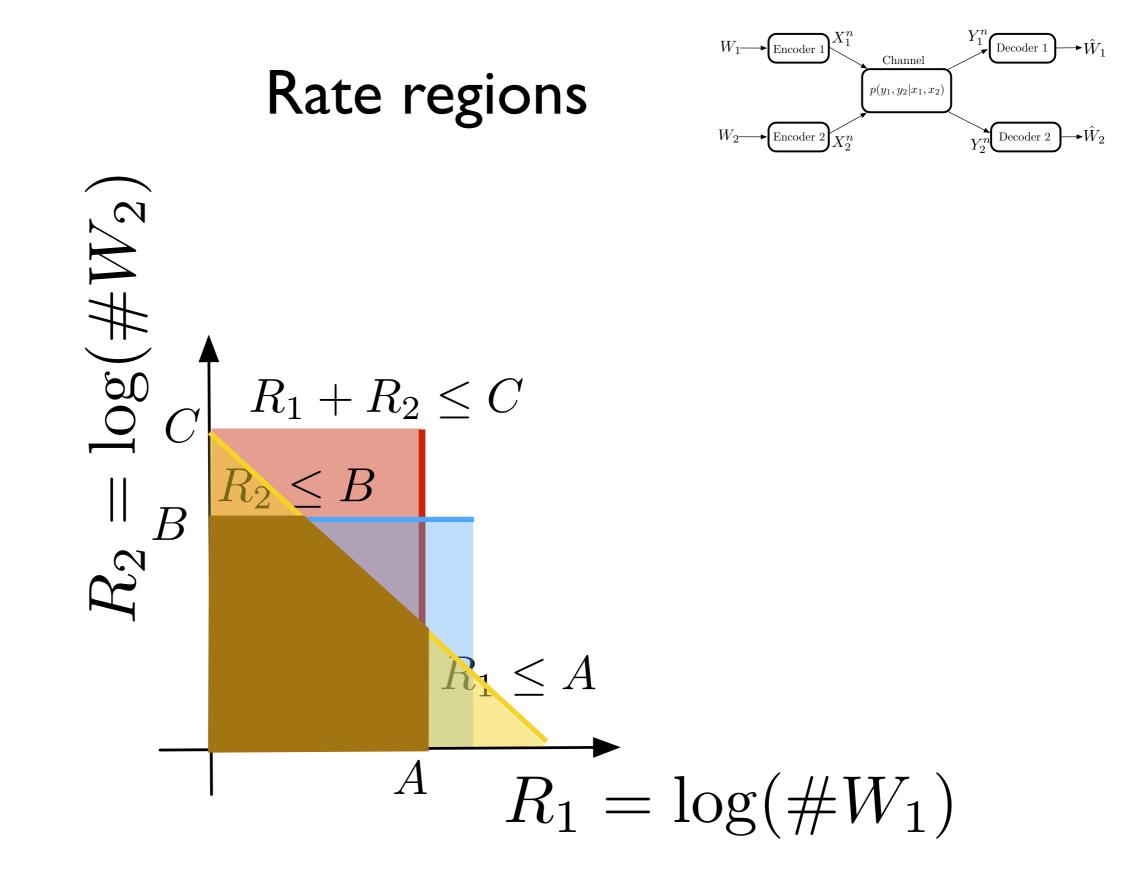
• Note: capacity region depends on $p(y_1, y_2 | x_1, x_2)$ only through the marginals $p(y_1 | x_1, x_2)$ and $p(y_2 | x_1, x_2)$

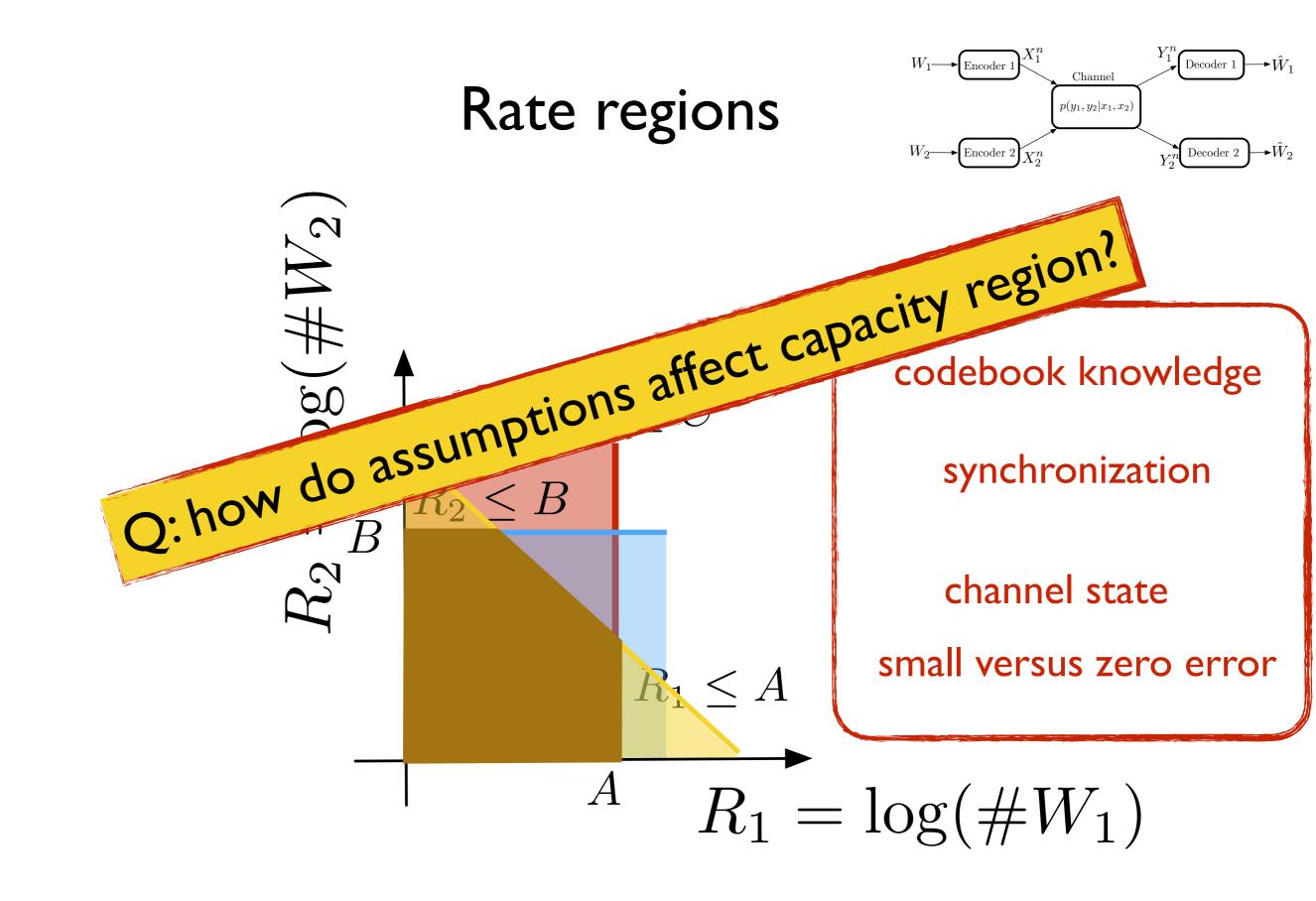








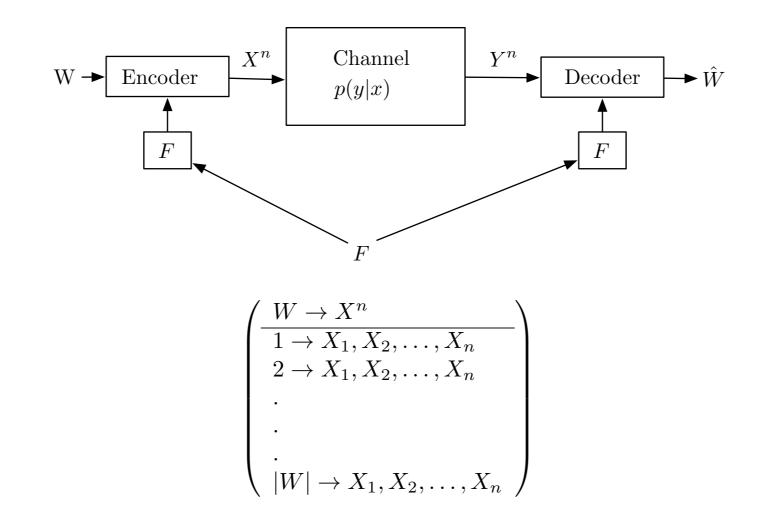




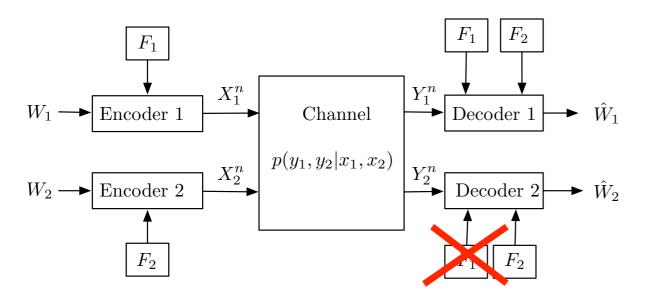
Codebook knowledge

Point to point codebook knowledge

"F" is the codebook, known to the Tx,Rx

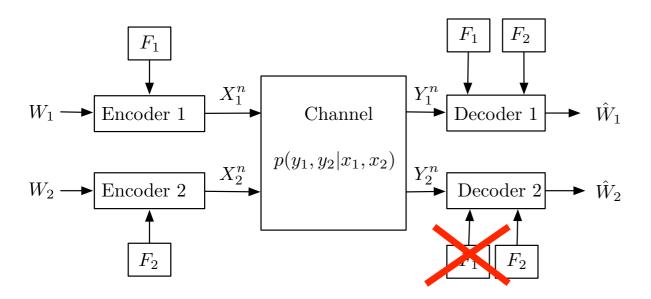


Our motivation:



IC with one oblivious Rx

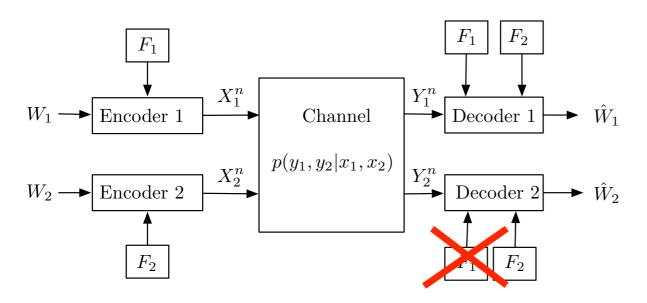
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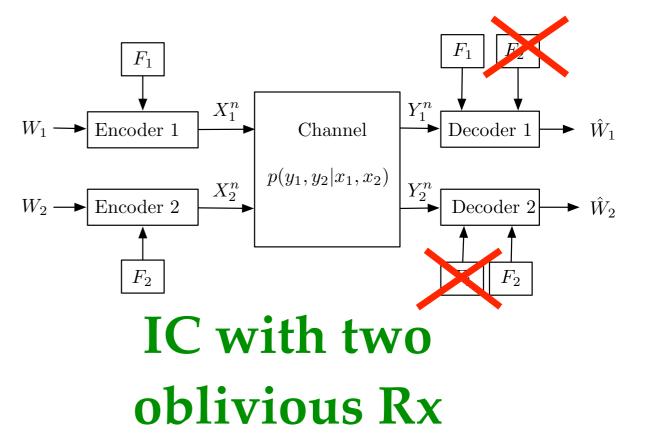
"primary user" in a cognitive setup

Our motivation:

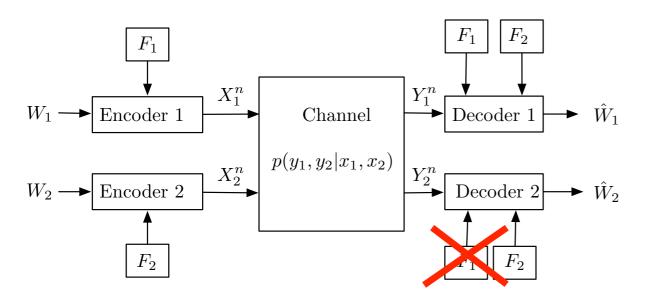


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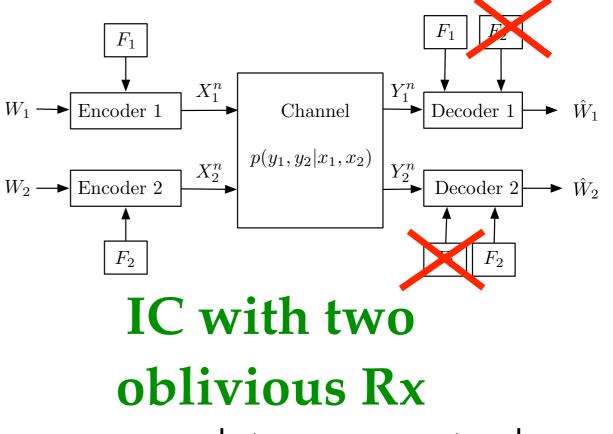


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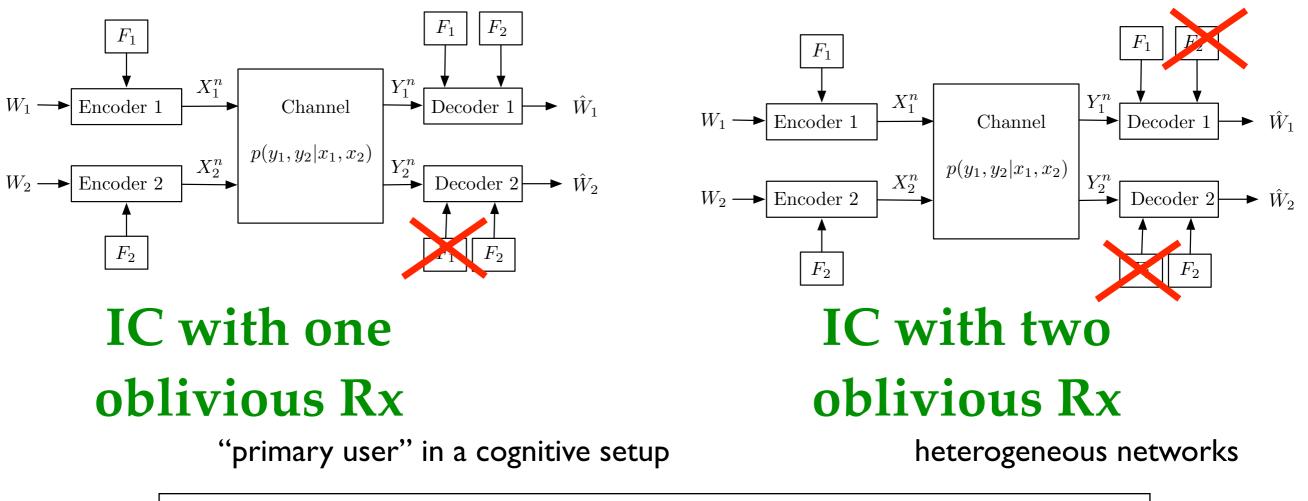
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"primary user" in a cognitive setup



heterogeneous networks

Our motivation:



A. Dytso, N. Devroye, and D. Tuninetti, "On the capacity of interference channels with partial codebook knowledge," ISIT 2013

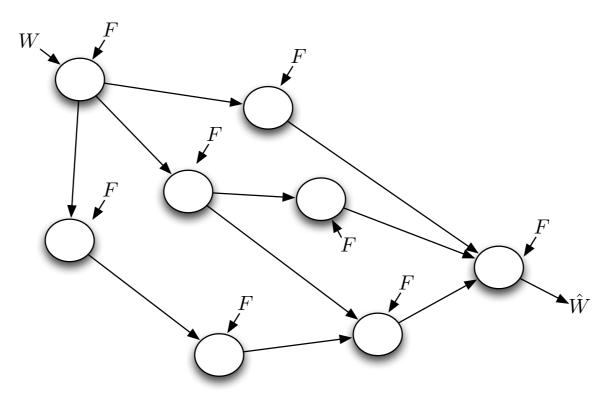
A. Dytso, D. Tuninetti and N. Devroye, ``On the Two-User Interference Channel With Lack of Knowledge of the Interference Codebook at One Receiver," IEEE Trans. on Infor. Theory, March 2015.

A. Dytso, D. Tuninetti and N. Devroye. "On Gaussian Interference Channels with Mixed Gaussian and Discrete Inputs," ISIT 2014

A. Dytso, D. Tuninetti and N. Devroye "Interference as Noise: Friend of Foe?" IEEE Trans. on Info Theory, June 2016.

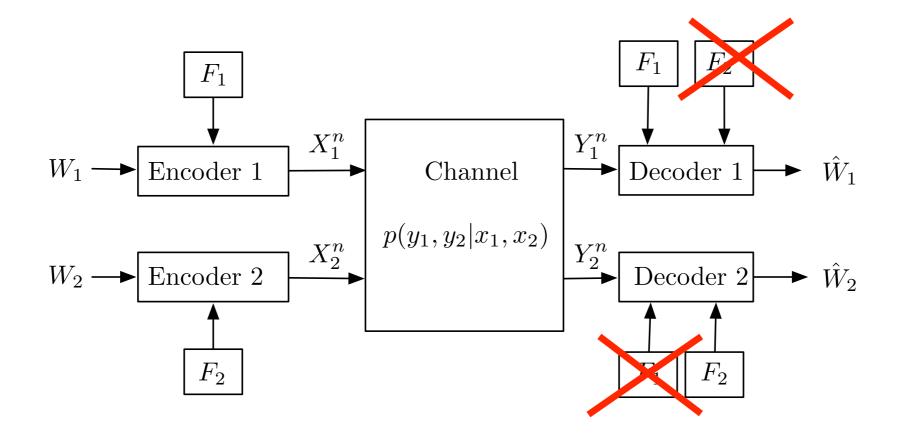
Networks with lack of codebook knowledge

 in networks, often assume nodes know all codebooks of ALL other nodes



•this may be unrealistic sometimes....

The interference channel



How do we use the codebook knowledge?

Interfering codebook knowledge in AWGN IC

• Use it to decode interference in very strong and strong interference regimes

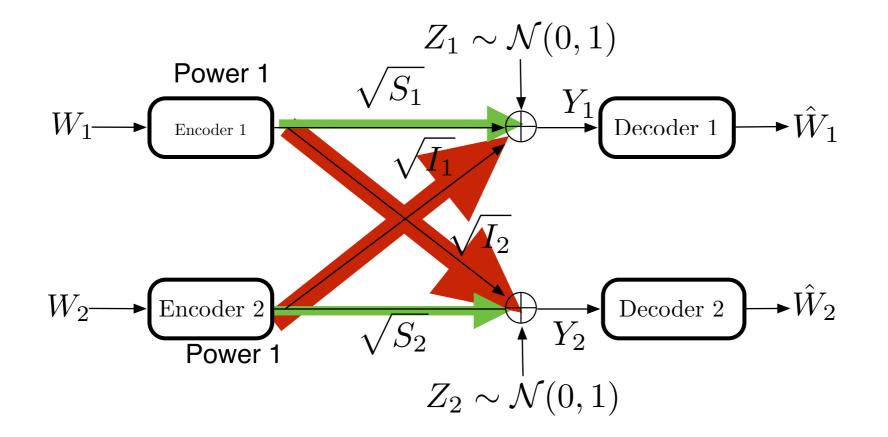
joint decoding, successive interference cancellation

A DM-IC is said to have very strong interference if

$$I(X_1; Y_1 | X_2) \le I(X_1; Y_2)$$

$$I(X_2; Y_2 | X_1) \le I(X_2; Y_1)$$

 $\forall p(x_1)p(x_2)$

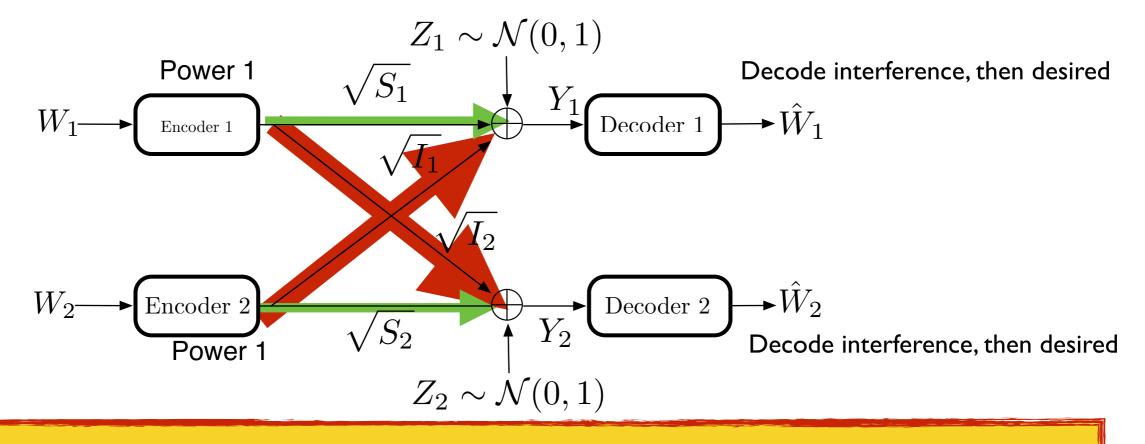


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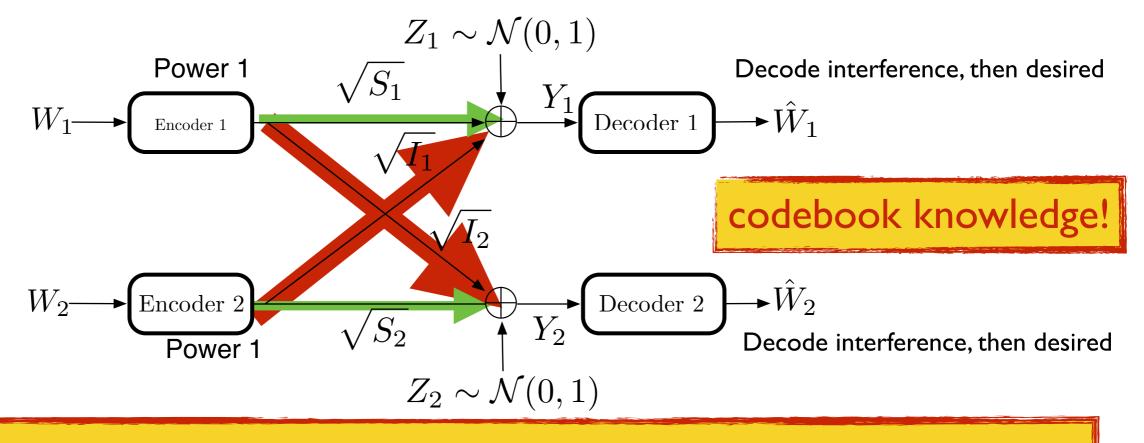
Successive interference cancellation achieves capacity!

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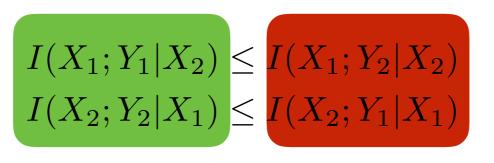
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$\forall p(x_1)p(x_2)$

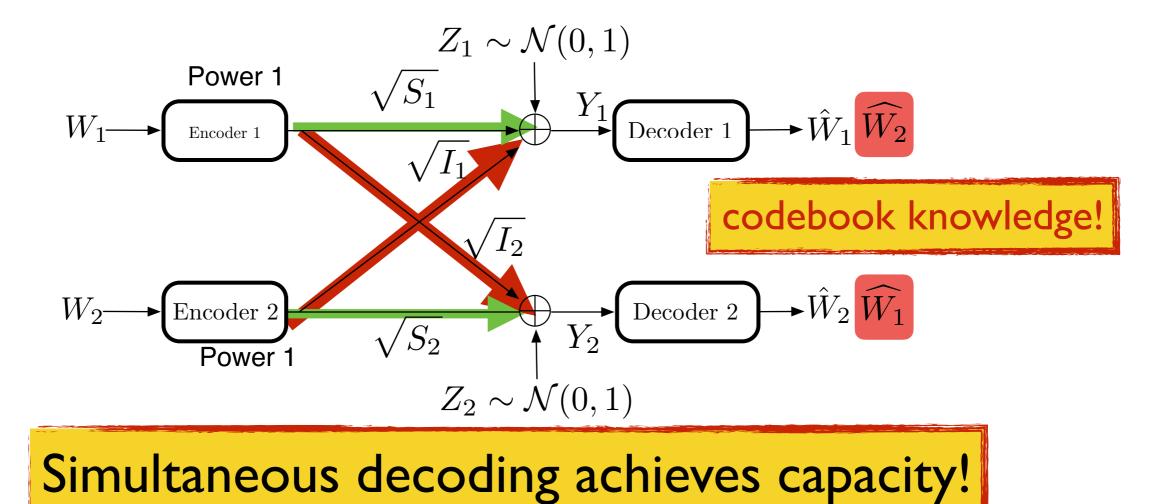


Successive interference cancellation achieves capacity!

A DM-IC is said to have strong interference if



 $\forall p(x_1)p(x_2)$



Interfering codebook knowledge in AWGN IC

• Use it to decode interference in very strong and strong interference regimes

Best general rate region

 Use it to decode public messages in Han + Kobayashi achievable rate region

+

[T. S. Han and K. Kobayashi, "A new achievable rate region for the interference channel," IEEE Trans. Inf. Theory, vol. 27, no. 1, pp. 49–60, 1981.]

Theorem (Han+Kobayashi inner bound). A rate pair (R_1, R_2) is achievable for a DM-IC $(\mathcal{X}_1 \times \mathcal{X}_2, p(y_1, y_2 | x_1, x_2), \mathcal{Y}_1 \times \mathcal{Y}_2)$ if it satisfies

 $\begin{aligned} R_{1} &\leq I(X_{1};Y_{1}|U_{2},Q) \tag{1} \\ R_{2} &\leq I(X_{2};Y_{2}|U_{1},Q) \tag{2} \\ R_{1} + R_{2} &\leq I(X_{1},U_{2};Y_{1}|Q) + I(X_{2};Y_{2}|U_{1},U_{2},Q) \tag{3} \\ R_{1} + R_{2} &\leq I(X_{1};Y_{1}|U_{1},U_{2},Q) + I(X_{2},U_{1};Y_{2}|Q) \tag{4} \\ R_{1} + R_{2} &\leq I(X_{1},U_{2};Y_{1}|U_{1},Q) + I(X_{2},U_{1};Y_{2}|U_{2},Q) \tag{5} \\ 2R_{1} + R_{2} &\leq I(X_{1},U_{2};Y_{1}|Q) + I(X_{1};Y_{1}|U_{1},U_{2},Q) + I(X_{2},U_{1};Y_{2}|U_{2},Q) \tag{6} \\ R_{1} + 2R_{2} &\leq I(X_{2},U_{1};Y_{2}|Q) + I(X_{2};Y_{2}|U_{1},U_{2},Q) + I(X_{1},U_{2};Y_{1}|U_{1},Q) \end{aligned}$

for some $p(q, u_1, u_2, x_1, x_2) = p(q)p(u_1, x_1|q)p(u_2, x_2|q)$ where $|\mathcal{U}_1| \le |\mathcal{X}_1| + 4$, $|\mathcal{U}_2| \le |\mathcal{X}_2| + 4$, and $|\mathcal{Q}| \le 7$.





[T. S. Han and K. Kobayashi, "A new achievable rate region for the interference channel," IEEE Trans. Inf. Theory, vol. 27, no. 1, pp. 49-60, 1981.]





Largest single-letter achievable rate region for IC

[T. S. Han and K. Kobayashi, "A new achievable rate region for the interference channel," IEEE Trans. Inf. Theory, vol. 27, no. 1, pp. 49-60, 1981.]

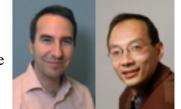
Achieves capacity when we know it

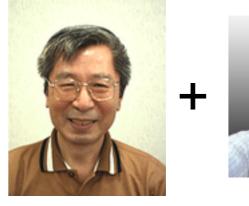
[H.-F. Chong, M. Motani, H. K. Garg, and H. El Gamal, "On the Han–Kobayashi region for the interference channel," IEEE Trans. Inf. Theory, vol. 54, no. 7, pp. 3188–3195, July 2008.]



(class of deterministic channels, approximately for class of semi-deterministic channels)

[A. El Gamal and M. H. M. Costa, "The capacity region of a class of deterministic interference channels," IEEE Trans. Inf. Theory, 1982.] .[I. E. Telatar and D. N. C. Tse, "Bounds on the capacity region of a class of interference channels," ISIT 2007]







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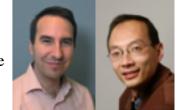
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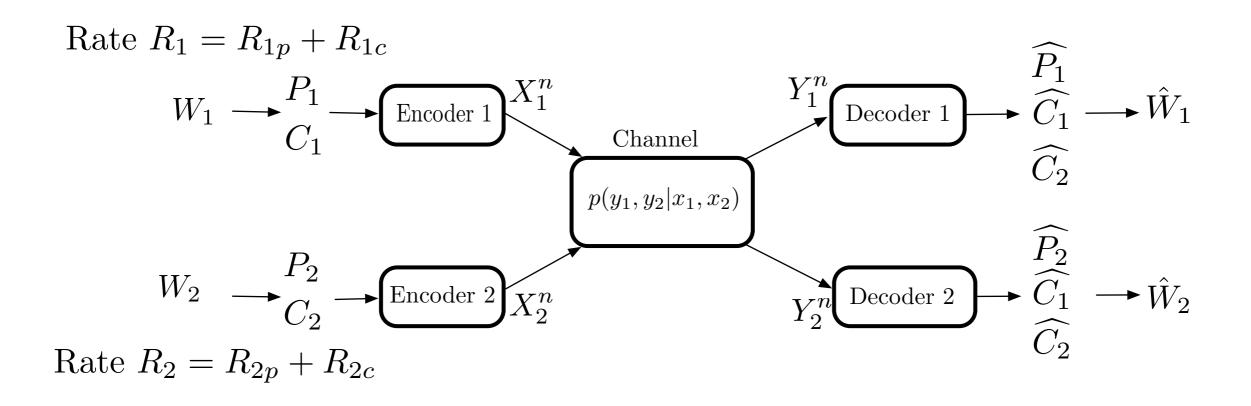


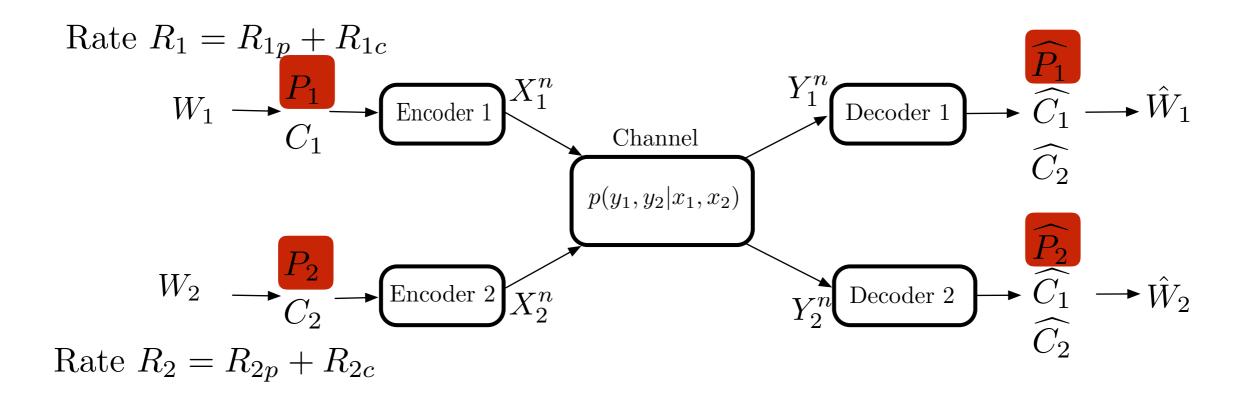
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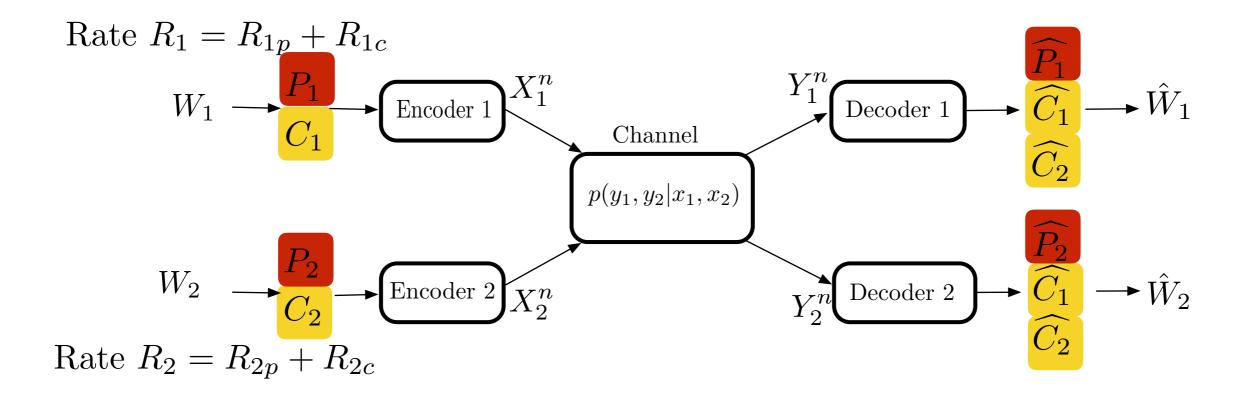


Requires knowledge of codebooks at both receivers - WHY?



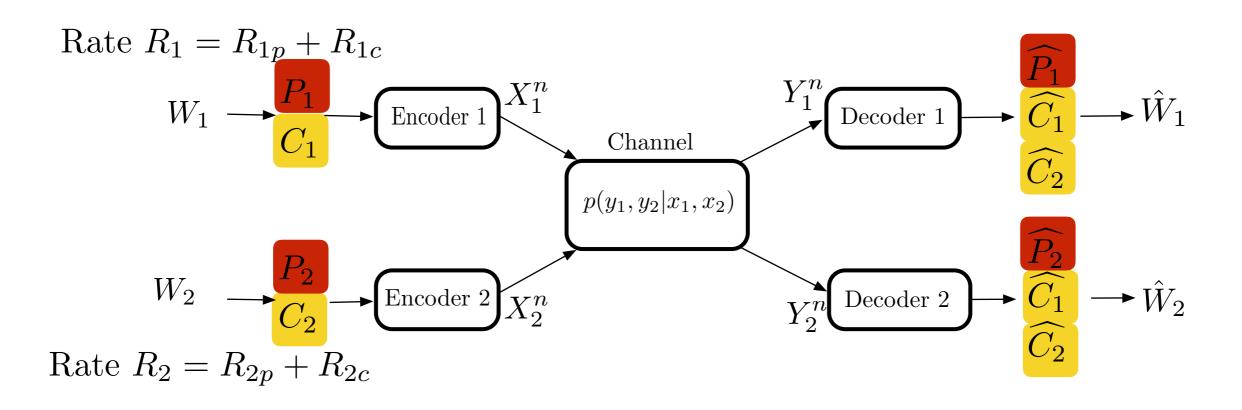


"Private" = decoded only by intended



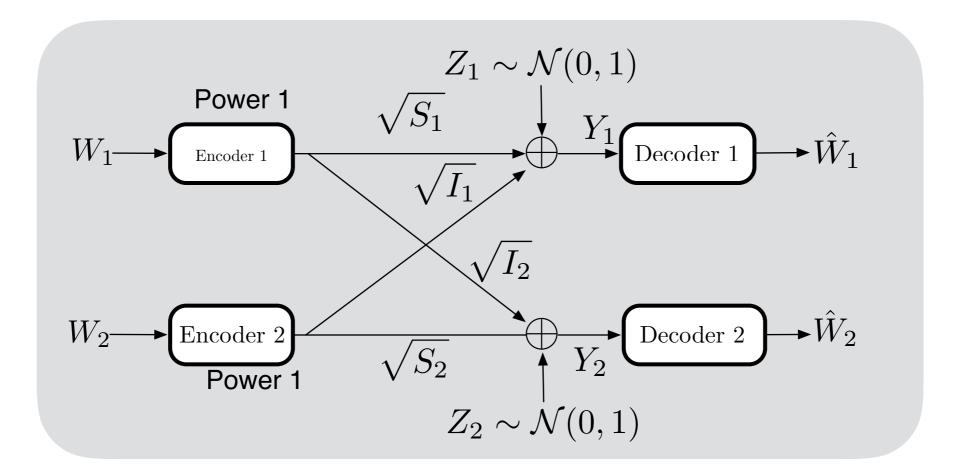
"Private" = decoded only by intended "Public" = common = decoded by everyone

requires interfering codebooks!



"Private" = decoded only by intended "Public" = common = decoded by everyone

The AWGN-IC



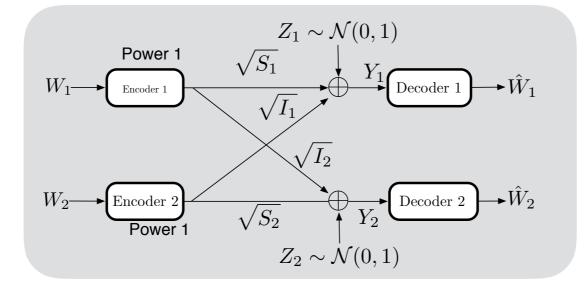
of practical relevance in wireless systems: cellular, wireless local area networks (WiFi), ad hoc networks (wireless sensors or nodes)

Interfering codebook knowledge in AWGN IC

• Use it to decode interference in very strong and strong interference regimes

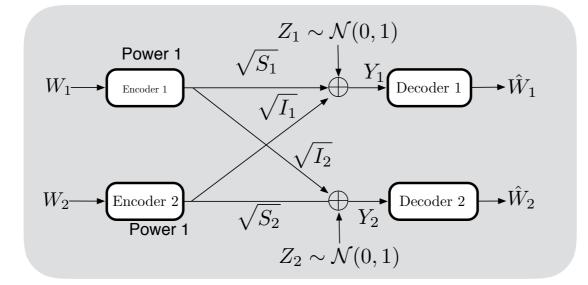
- Use it to decode public messages in Han + Kobayashi achievable rate region
- Use it to achieve capacity to within 1/2 bit for Gaussian noise channels

AWGN: H+K achieves capacity to within 1/2 bit



Theorem (gap for Gaussian IC) If (R_1, R_2) is in the outer bound \mathcal{R}_O^{AWGN} then $(R_1 - 1/2, R_2 - 1/2)$ is achievable.

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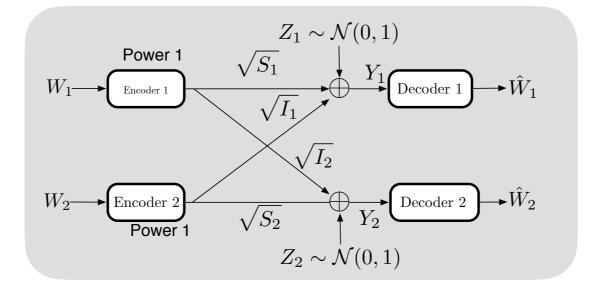


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Etkin, Tse, Wang show how to pick Gaussian inputs in H+K scheme

[R. Etkin, D. Tse, and H. Wang, "Gaussian interference channel capacity to within one bit," IEEE Trans. Inf. Theory, vol. 54, no. 12, pp. 5534–5562, Dec. 2008.]

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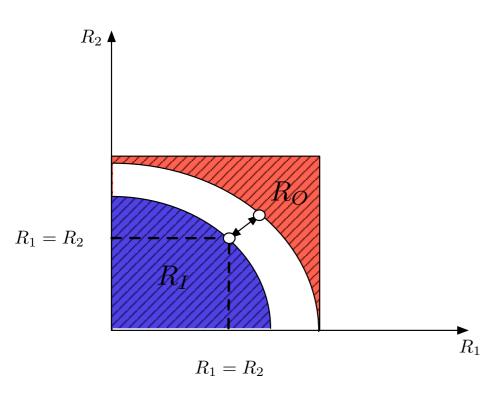
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depends on the regime of operation

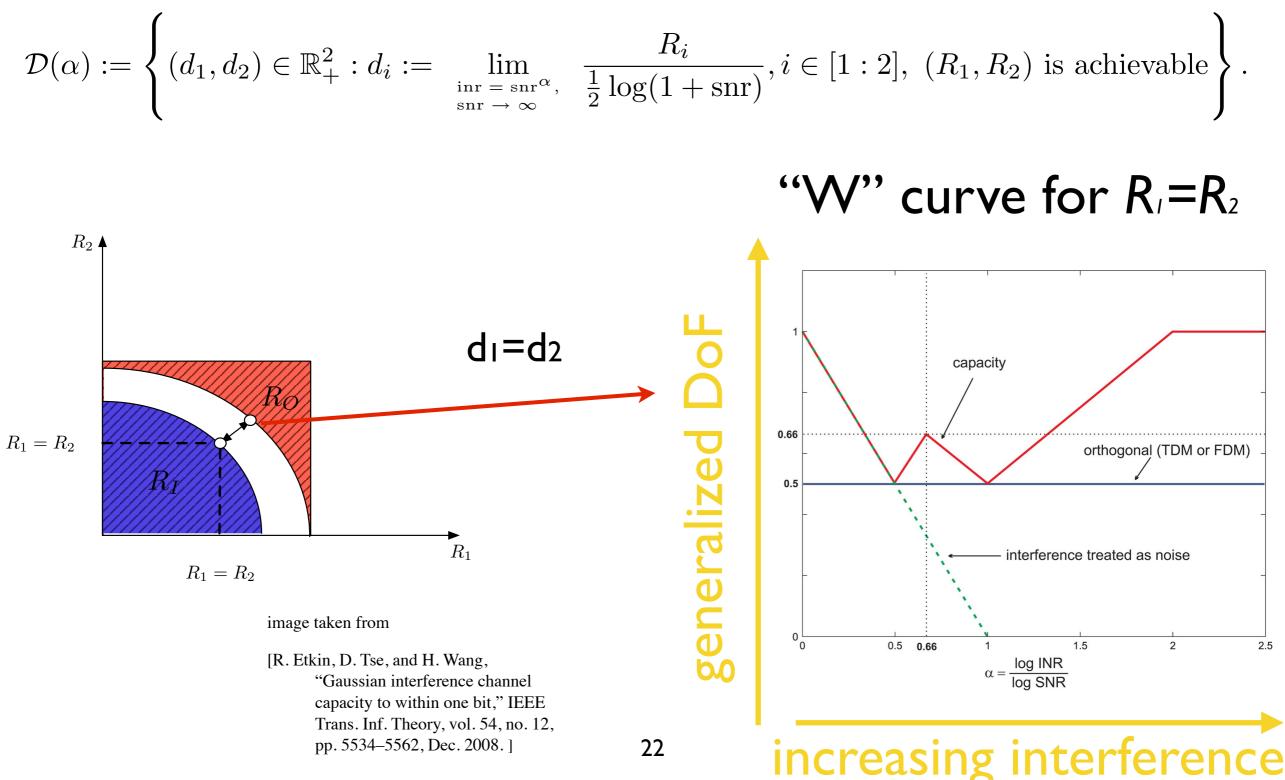
AWGN: the "W" curve for the GDoF

highlights effect of interference rather than noise

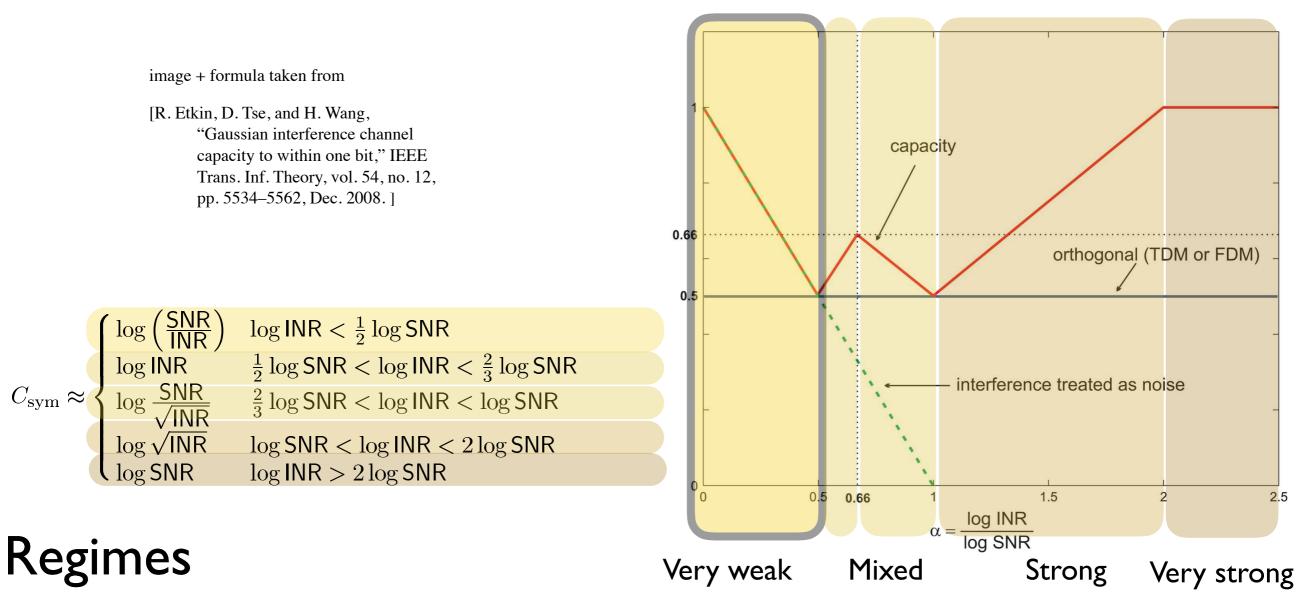
$$\mathcal{D}(\alpha) := \left\{ (d_1, d_2) \in \mathbb{R}^2_+ : d_i := \lim_{\substack{\inf = \operatorname{snr}^\alpha, \\ \operatorname{snr} \to \infty}} \frac{R_i}{\frac{1}{2} \log(1 + \operatorname{snr})}, i \in [1:2], \ (R_1, R_2) \text{ is achievable} \right\}.$$



highlights effect of interference rather than noise

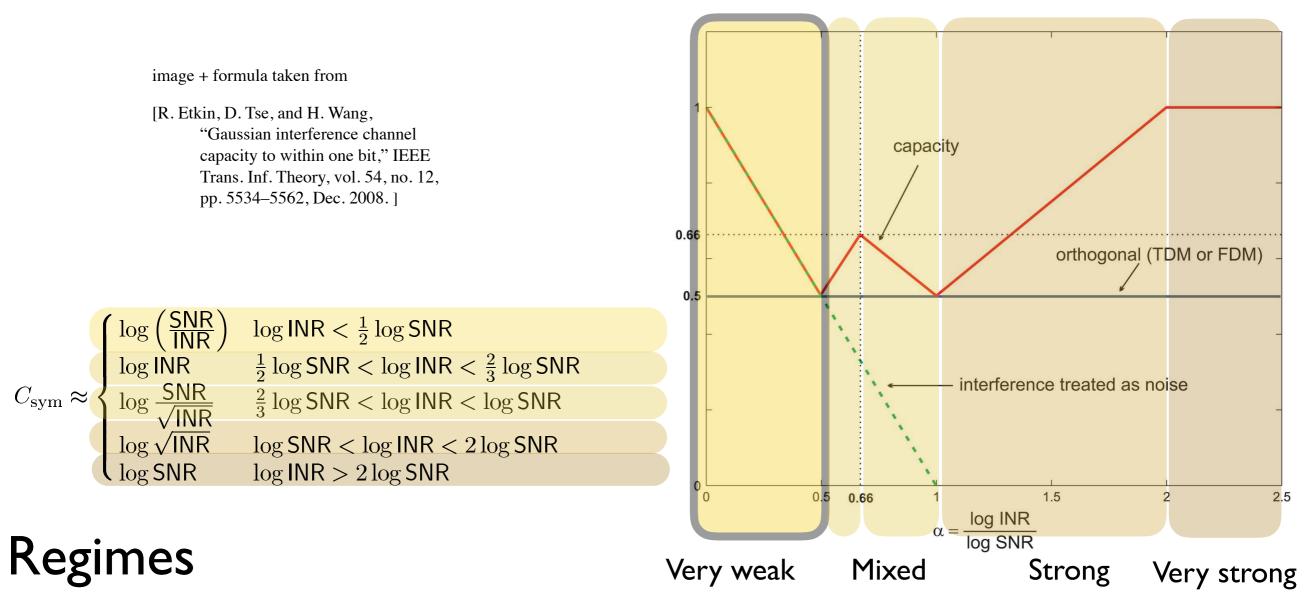


2.5



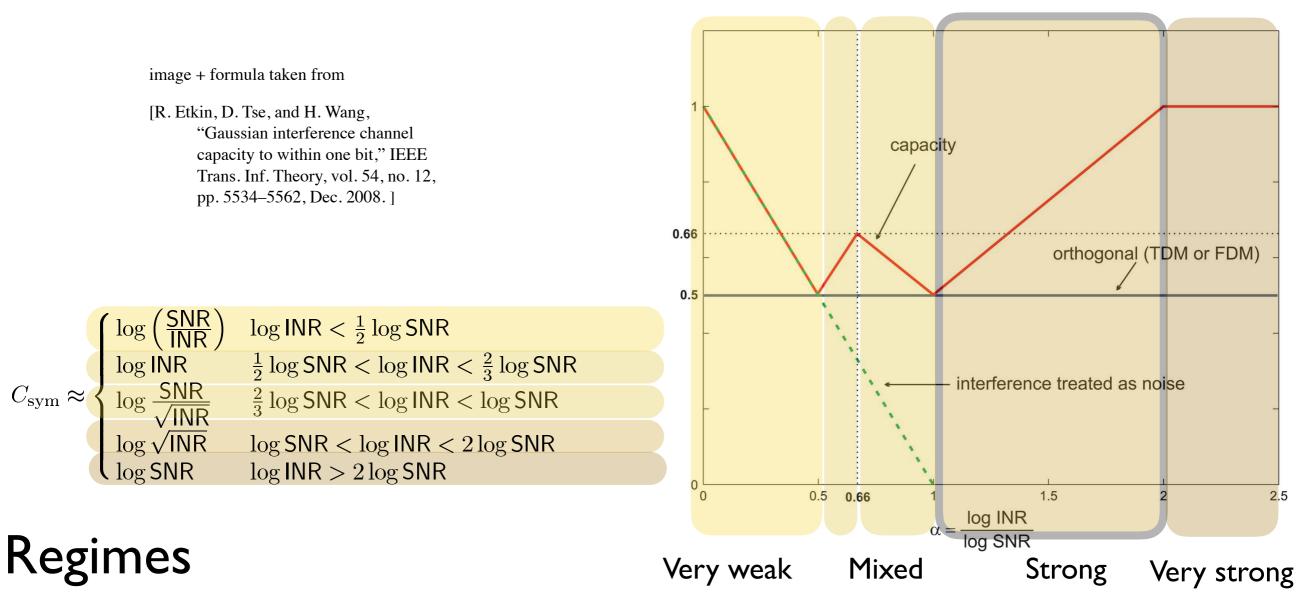
Very weak: treating interference as noise is sometimes sum-capacity optimal, capacity partially known

- [X. Shang, G. Kramer, and B. Chen, "A new outer bound and the noisy-interference sum-rate capacity for Gaussian interference channels," IEEE Trans. Inf. Theory, vol. 55, no. 2, pp. 689–699, Feb. 2009.]
- [V. S. Annapureddy and V. V. Veeravalli, "Gaussian interference networks: Sum capacity in the low interference regime and new outer bounds on the capacity region," IEEE Trans. Inf. Theory, vol. 55, no. 7, pp. 3032–3050, July 2009.]
- [A. S. Motahari and A. K. Khandani, "Capacity bounds for the Gaussian interference channel," IEEE Trans. Inf. Theory, vol. 55, no. 2, pp. 620–643, Feb. 2009.]



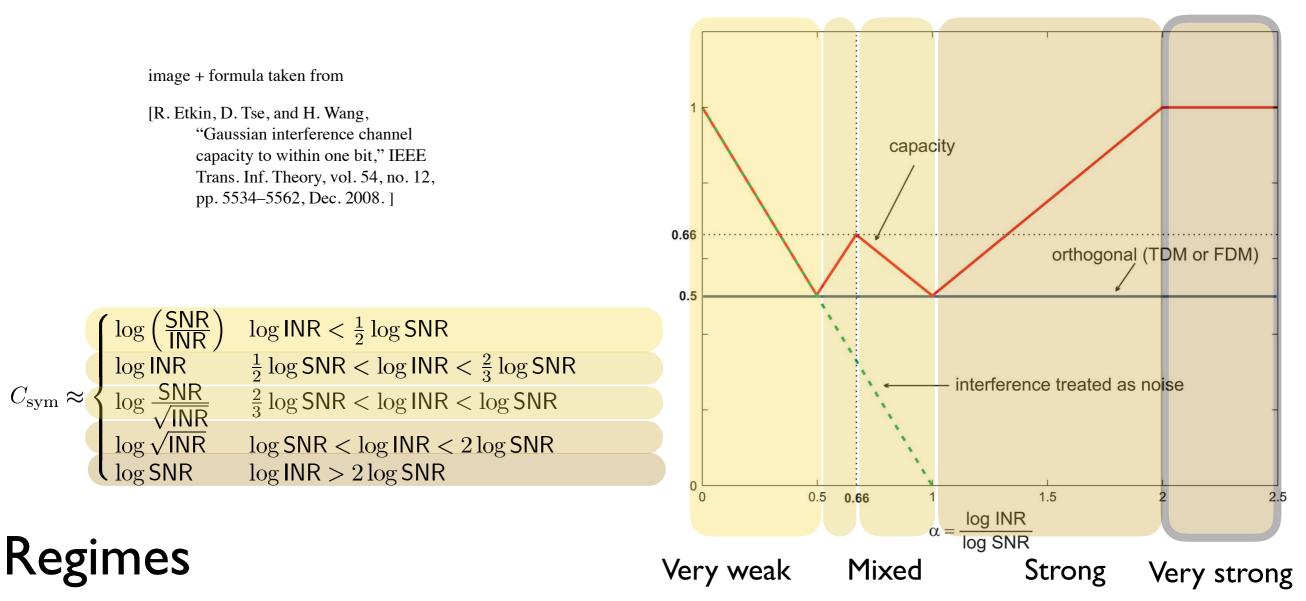
Very weak: treating interference as noise is sometimes sum-capacity optimal, capacity partially known

Treating interference as noise inner bound (with Gaussian inputs): $R_1 \leq \frac{1}{2} \log \left(1 + \frac{S_1}{1 + I_1} \right), \quad R_2 \leq \frac{1}{2} \log \left(1 + \frac{S_2}{1 + I_2} \right)$



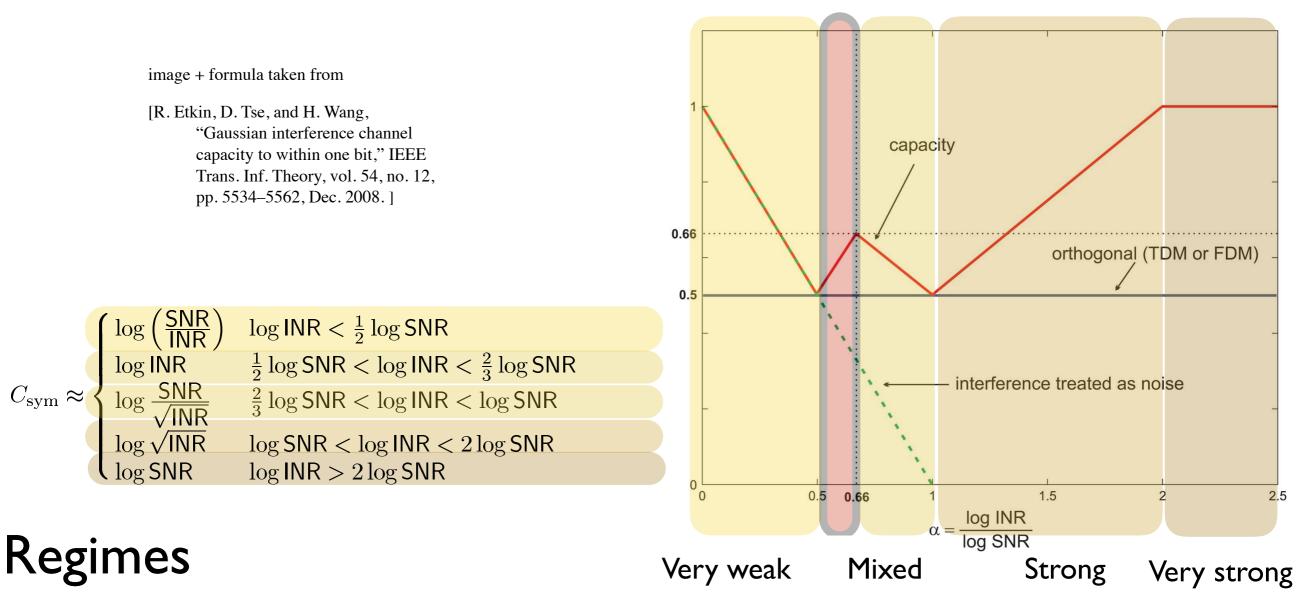
Very weak: treating interference as noise is sometimes sum-capacity optimal, capacity partially known

Strong: jointly decoding both messages at both receivers is capacity optimal, capacity known



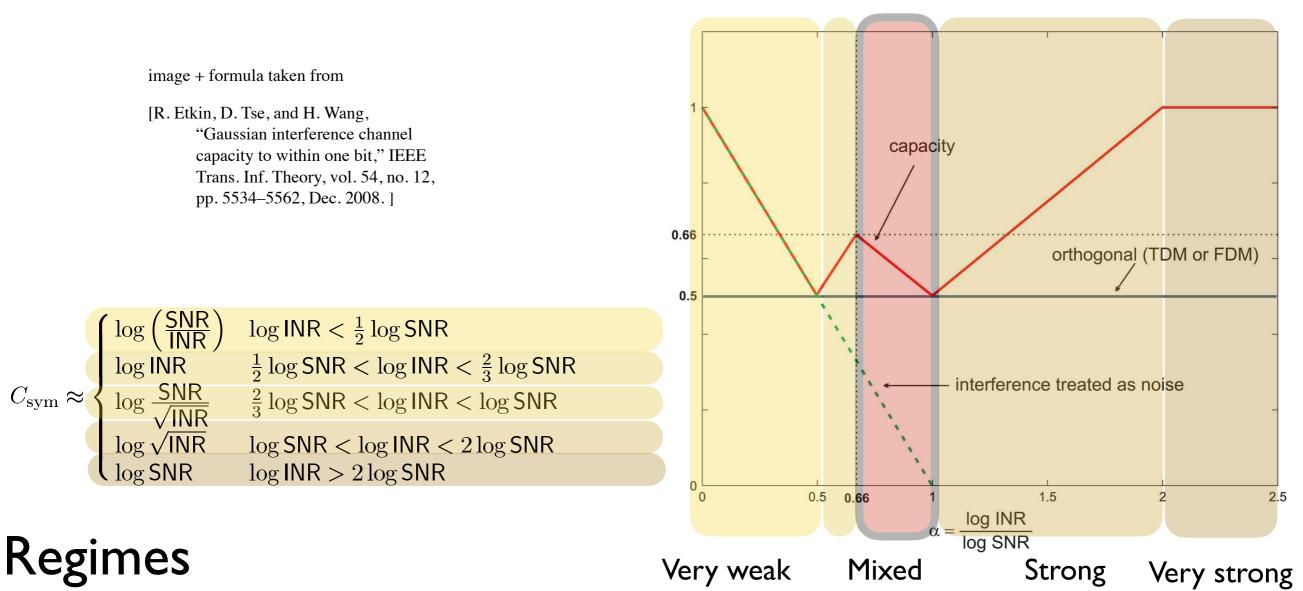
Very weak: treating interference as noise is sometimes sum-capacity optimal, capacity partially known

Strong: jointly decode both messages at both receivers is capacity optimal, capacity known Very strong: first decode interference then desired is capacity optimal, capacity known

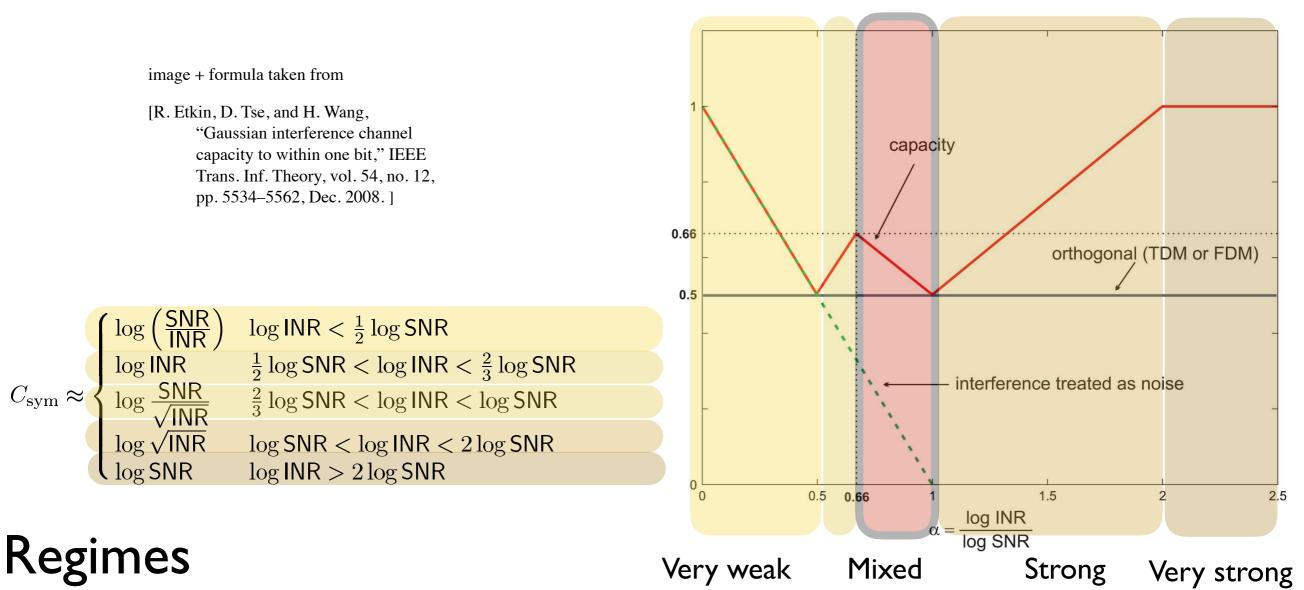


Very weak: treating interference as noise is sometimes capacity optimal, capacity partially known Mixed I: partially decode interference H+K is gDoF optimal — larger INR, cancel more, capacity unknown

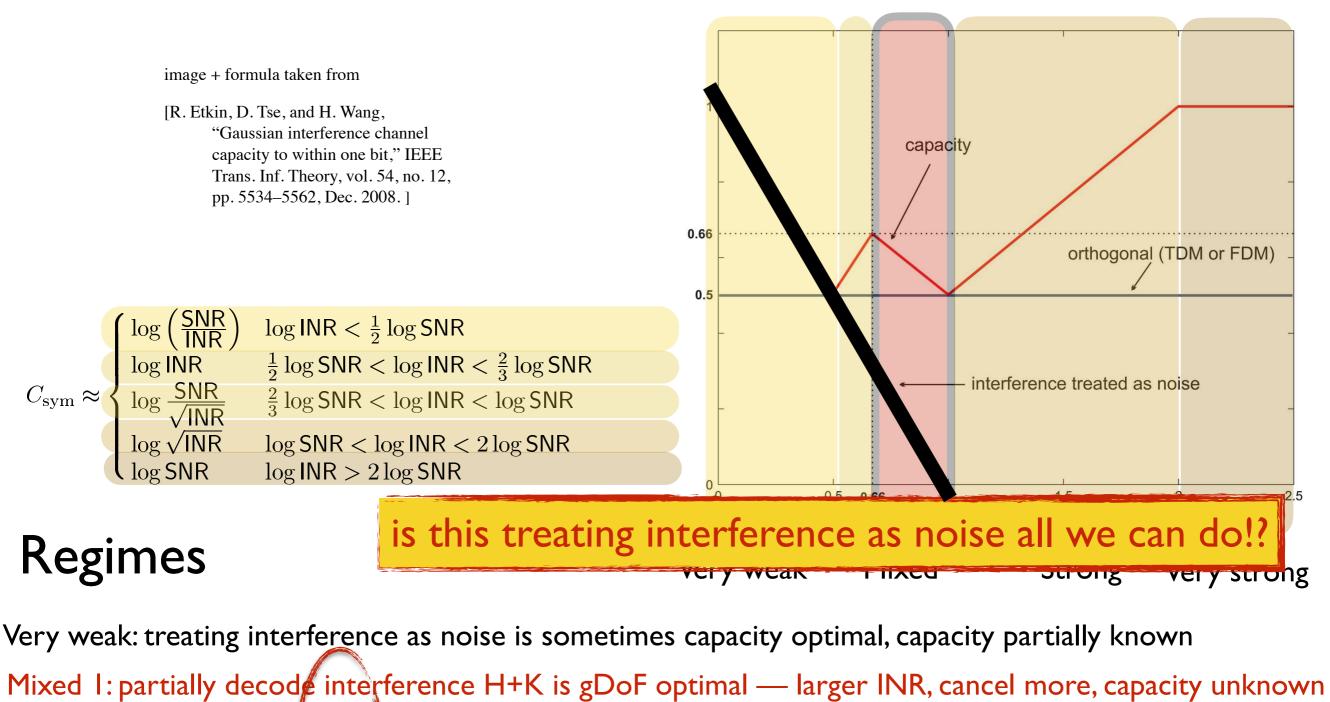
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Very weak: treating interference as noise is sometimes capacity optimal, capacity partially known Mixed I: partially decode interference H+K is gDoF optimal — larger INR, cancel more, capacity unknown Mixed 2: partially decode interference H+K is gDoF optimal — larger INR hurts, capacity unknown Strong: jointly decode both messages at both receivers is capacity optimal, capacity known Very strong: first decode interference then desired is capacity optimal, capacity known 27



Very weak: treating interference as noise is sometimes capacity optimal, capacity partially known Mixed 1: partially decode interference H+K is gDoF optimal — larger INR, cancel more, capacity unknown Mixed 2: partially decode interference H+K is gDoF optimal — larger INP, burte, consist unknown Strong: jointly decode interference then desired is capacity optimal, capacity known Very strong: first decode interference then desired is capacity optimal, capacity known 27



Mixed 2: partially decode into Korence UUK is sDeE optimal Jargor INP, burte exercise unleave Strong: jointly decode interfering codebooks to (partially) decode interference! Very strong: first decode interference then desired is capacity optimal, capacity known 27



Back to the question



Daniela Tuninetti

Natasha Devroye

Alex Dysto

How does lack of codebook knowledge affect capacity of the Gaussian IC?

some slides taken from Alex Dytso's Ph.D. defense, May 2016

Idea: use non-Gaussian inputs in a Gaussian interference channel!



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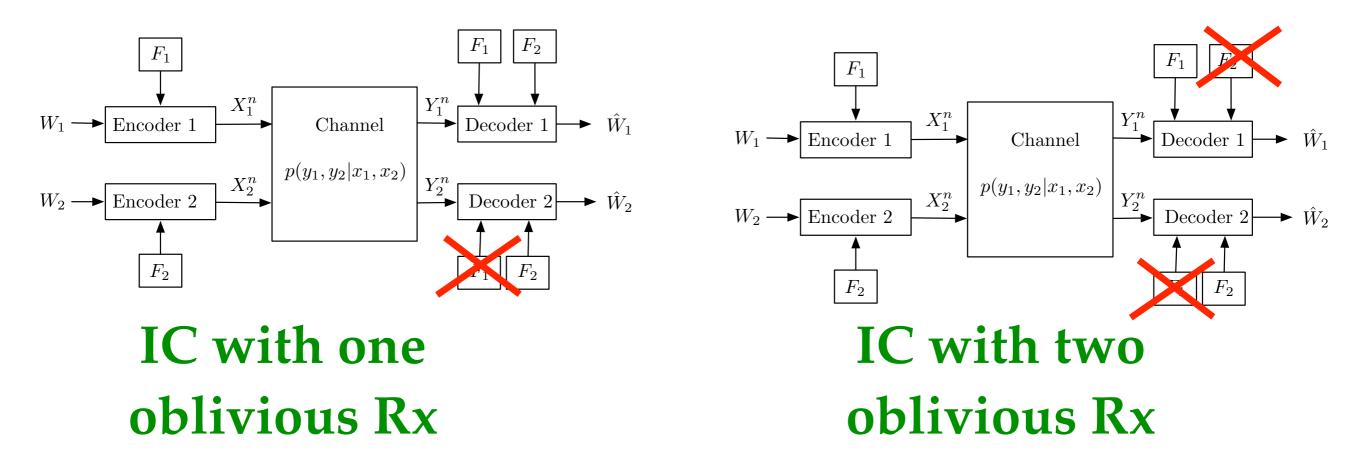
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Idea: use non-Gaussian inputs in a Gaussian interference channel!

Decent performance, and can``estimate" and strip off!

ICs with lack of codebook knowledge

Our motivation:



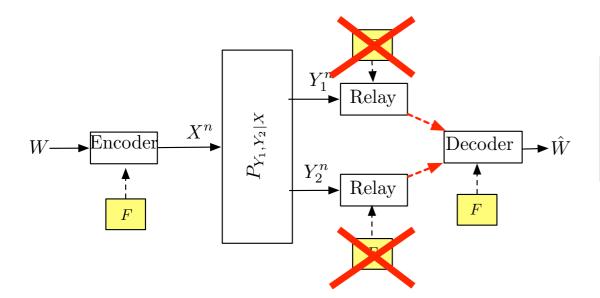
A. Dytso, N. Devroye, and D. Tuninetti, "On the capacity of interference channels with partial codebook knowledge," ISIT 2013

A. Dytso, D. Tuninetti and N. Devroye, ``On the Two-User Interference Channel With Lack of Knowledge of the Interference Codebook at One Receiver," IEEE Transactions on Information Theory, Vol. 61, No. 3, pp. 1256-1276, March 2015.

A. Dytso, D. Tuninetti and N. Devroye. "On Gaussian Interference Channels with Mixed Gaussian and Discrete Inputs," ISIT 2014

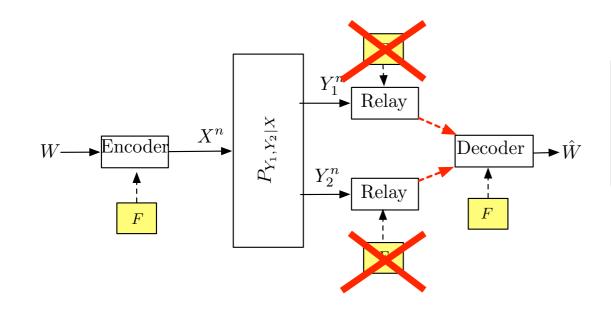
A. Dytso, D. Tuninetti and N. Devroye "Interference as Noise: Friend of Foe?" IEEE Trans. on Info Theory, June 2016.

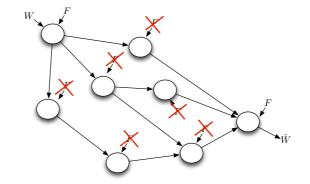
Past work: lack of codebooks leads to non-Gaussians outperforming Gaussians



- A. Sanderovich, S. Shamai, Y. Steinberg, and G. Kramer, "*Communication via decentralized processing*," IT July 2008.
- 1. Upper and lower bounds, which coincide for deterministic channels
- 2. Gaussian noise: optimizing input unknown
- 3. Gaussian noise: example where BPSK outperforms Gaussian inputs

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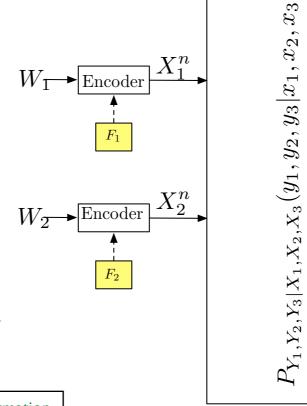
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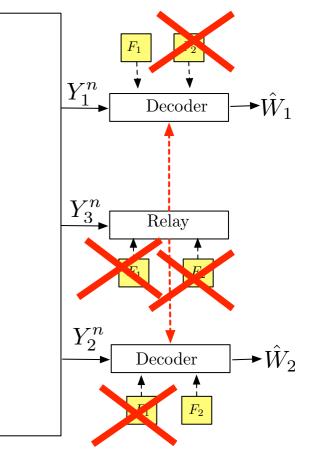
O. Simeone, E. Erkip, and S. Shamai, "*On codebook information for interference relay channels with out-of-band relaying*," IT May 2011.

- 1. Primitive relay channel: capacity with compress forward
- 2. IC+R+Oblivious receivers: capacity with compress forward and TIN
- 3. Gaussian noise: optimizing input unknown

$$\mathcal{C}^{\text{IC-OR}} = \bigcup_{P_Q P_{X_1|Q} P_{X_2|Q}} \left\{ \begin{array}{l} R_1 \le I(X_1; Y_1|Q) \\ R_2 \le I(X_2; Y_2|Q) \end{array} \right\}$$

[Ye Tian and Aylin Yener, Relaying for Multiuser Networks in the Absence of Codebook Information, IEEE Transactions on Information Theory, 61(3), pp. 1247-1256, Mar. 2015.]





Discrete inputs in Gaussian channels — deeper?

Other supporting arguments

- E. Abbe and L. Zheng, "*A coordinate system for Gaussian networks*," IT 2012.
- E. Calvo, J. Fonollosa, and J. Vidal, "On the totally asynchronous interference channel with single-user receivers," ISIT 2009

- No gDoF Gain
- Discrete input conclusions are simulation based

Questions

 loss in performance in IC due to lack of interfering codebook knowledge?

•are there inputs that outperform Gaussians in the AWGN IC under these conditions?

• can we show analytical gains?

best inner bound for Gaussian IC is the complex H+K scheme

- best inner bound for Gaussian IC is the complex H+K scheme
- simpler scheme Treating Interference as Noise with no Time Sharing:

$$\mathcal{R}_{\text{in}}^{\text{TINnoTS}} = \bigcup_{P_{X_1 X_2} = P_{X_1} P_{X_2}} \left\{ \begin{array}{l} 0 \le R_1 \le I(X_1; Y_1) \\ 0 \le R_2 \le I(X_2; Y_2) \end{array} \right\}$$

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• we show discrete inputs in TINnoTS performs well!

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- we show discrete inputs in TINnoTS performs well!
- neat, general tools to bound minimum distance of sum-sets, and mutual information achieved by discrete RVs in Gaussian noise along the way

Similar results as

S. Li, Y.-C. Huang, T. Liu, and H.D. Pfister, "On the limits of treating interference as noise in the two-user Gaussian symmetric interference channel," ISIT 2015.

Capacity:
$$C = \lim_{n \to \infty} co\left(\bigcup_{P_{X_1^n X_2^n} = P_{X_1^n} P_{X_2^n}} \left\{ \begin{array}{l} 0 \le R_1 \le \frac{1}{n} I(X_1^n; Y_1^n) \\ 0 \le R_2 \le \frac{1}{n} I(X_2^n; Y_2^n) \end{array} \right\} \right)$$

R. Ahlswede, "Multi-way communication channels," in Proc. IEEE Int. Symp. Inf. Theory, March 1973, pp. 23–52.

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i.i.d. inputsTreat Interference as Noise Inner Bound: $\mathcal{R}_{in}^{\text{TIN+TS}} = co\left(\bigcup_{P_{X_1X_2}=P_{X_1}P_{X_2}} \left\{ \begin{array}{l} 0 \leq R_1 \leq I(X_1;Y_1) \\ 0 \leq R_2 \leq I(X_2;Y_2) \end{array} \right\} \right)$ With Time Sharing $\mathcal{R}_{in}^{\text{TINnoTS}} = \bigcup_{P_{X_1X_2}=P_{X_1}P_{X_2}} \left\{ \begin{array}{l} 0 \leq R_1 \leq I(X_1;Y_1) \\ 0 \leq R_2 \leq I(X_2;Y_2) \end{array} \right\}$ No Time Sharing

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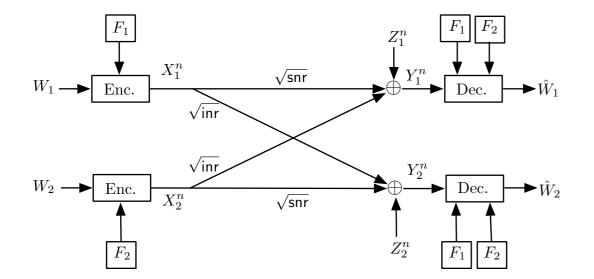
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i.i.d. inputs Treat Interference as Noise Inner Bound:

 $\begin{array}{ll} \mathcal{R}_{\text{in}}^{\text{TIN+TS}} &= \operatorname{co} \left(\bigcup_{P_{X_1X_2} = P_{X_1}P_{X_2}} \left\{ \begin{array}{l} 0 \leq R_1 \leq I(X_1;Y_1) \\ 0 \leq R_2 \leq I(X_2;Y_2) \end{array} \right\} \right) \text{ With Time Sharing} \\ \\ \mathcal{R}_{\text{in}}^{\text{TINnoTS}} &= \bigcup_{P_{X_1X_2} = P_{X_1}P_{X_2}} \left\{ \begin{array}{l} 0 \leq R_1 \leq I(X_1;Y_1) \\ 0 \leq R_2 \leq I(X_2;Y_2) \end{array} \right\} & \text{ No Time Sharing} \\ \\ & \text{ How far away is TINnoTS from capacity?} \\ \\ & \text{ Is it really "treating interference as noise"?} \end{array} \end{array}$

Gaussian channels with discrete inputs



$$Z_1, Z_2 \sim \mathcal{N}(0, 1)$$

$$Y_1 = \sqrt{\operatorname{snr}} X_1 + \sqrt{\operatorname{inr}} X_2 + Z_1$$
$$Y_2 = \sqrt{\operatorname{inr}} X_1 + \sqrt{\operatorname{snr}} X_2 + Z_2$$

- instead of taking X1 and X2 to be Gaussian, take them to be discrete
- difficulty: how to evaluate mutual information expressions with discrete and Gaussian mixtures

$$\mathcal{R}_{\text{in}}^{\text{TINnoTS}} = \bigcup_{P_{X_1 X_2} = P_{X_1} P_{X_2}} \left\{ \begin{array}{l} 0 \le R_1 \le I(X_1; Y_1) \\ 0 \le R_2 \le I(X_2; Y_2) \end{array} \right\}$$

Tools for Discrete Inputs

• Discrete input

$$X_D \sim P(X_D) = \sum_{i=1}^{|X|} p_i \delta(x_i)$$

• Discrete input

• PAM input

• Discrete input

• PAM input

• Minimum distance

$$d_{\min}(X_D) = \min_{x_i, x_j: i \neq j} \|x_i - x_i\|$$

- $X_D \sim P(X_D) = \sum_{i=1}^{|X|} p_i \delta(x_i)$ • Discrete input $X_D \sim \text{PAM}(N), |X| = N, p_i = \frac{1}{N} \text{ for all } i \in [1, ..., N]$ • PAM input x
- Minimum distance

$$d_{\min}(X_D) = \min_{x_i, x_j: i \neq j} \|x_i - x_i\|$$

Mixed inputs

$$X_{\min} = \sqrt{1 - \delta} X_D + \sqrt{\delta} X_G,$$

$$\delta \in [0, 1],$$

$$X_G \sim \mathcal{N}(0, 1)$$

$$\mathbb{E}[X_D^2] \le 1$$

Discrete input

• PAM input

Minimum distance

$$d_{\min}(X_D) = \min_{x_i, x_j: i \neq j} \|x_i - x_i\|$$

• Mixed inputs

$$X_{\text{mix}} = \sqrt{1 - \delta} X_D + \sqrt{\delta} X_G,$$

$$\delta \in [0, 1],$$

$$X_G \sim \mathcal{N}(0, 1)$$

$$\mathbb{E}[X_D^2] \le 1$$

Setup:

$$Y = \sqrt{\operatorname{snr}} X + Z,$$
$$Z \sim \mathcal{N}(0, 1)$$

We define:
$$I(X; Y_{snr}) = I(X, snr)$$

 $\mathbb{E} \left[(X - \mathbb{E}[X|Y_{snr}])^2 \right] = mmse(X, snr)$

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Interested in: $[H(X_D) - gap]^+ \le I(X_D, snr) \le H(X_D)$

Want the tightest version of the "gap" term for a given PMF

$$[H(X_D) - gap]^+ \le I(X_D, \operatorname{snr}) \le H(X_D)$$

Dzarow-Wyner-A
$$gap_{OW-A} \le \xi \log \frac{1}{\xi} + (1 - \xi) \log \frac{1}{1 - \xi} + \xi \log(N - 1), \ \xi := 2Q\left(\frac{\sqrt{\operatorname{snr}}d_{\min(X_D)}}{2}\right)$$

$$\begin{aligned} \mathbf{Ozarow-Wyner-B}\\ \mathsf{gap} &\leq \frac{1}{2}\log\left(\frac{\pi e}{6}\right) + \frac{1}{2}\log\left(1 + \frac{\mathrm{lmmse}(X,\mathsf{snr})}{d_{\min}(X_D)^2}\right) \end{aligned}$$

L. Ozarow and A. Wyner, "On the capacity of the Gaussian channel with a finite number of input levels," IEEE Trans. Inf. Theory, vol. 36, no. 6, pp. 1426–1428, Nov 1990.

$$[H(X_D) - gap]^+ \le I(X_D, \operatorname{snr}) \le H(X_D)$$

Ozarow-Wyner-A $\operatorname{gap}_{OW-A} \leq \xi \log \frac{1}{\xi} + (1 - \xi) \log \frac{1}{1 - \xi} + \xi \log(N - 1) , \ \xi := 2Q \left(\frac{\sqrt{\operatorname{snr}} d_{\min(X_D)}}{2} \right)$

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$$\left[-\log\left(\sum_{(i,j)\in[1:N]^2}\frac{p_ip_j}{\sqrt{4\pi}}e^{-\frac{\operatorname{snr}(x_i-x_j)^2}{4}}\right)-\frac{1}{2}\log\left(2\pi e\right)\right]^+ \leq I(X_D,\operatorname{snr}) \leq H(X_D)$$

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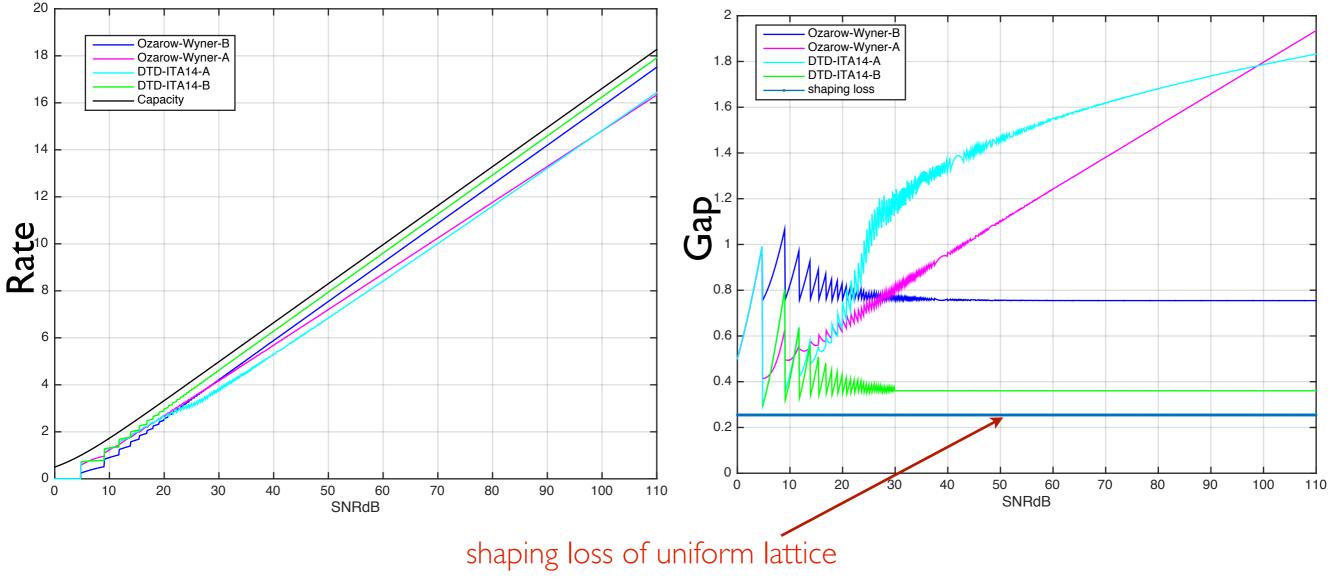
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$$\begin{aligned} \mathbf{DTD-ITA^{14-B}} \\ \left[-\log\left(\sum_{(i,j)\in[1:N]^2} \frac{p_i p_j}{\sqrt{4\pi}} e^{-\frac{\operatorname{snr}(x_i - x_j)^2}{4}}\right) - \frac{1}{2}\log\left(2\pi e\right) \right]^+ &\leq I(X_D, \operatorname{snr}) \leq H(X_D) \\ \operatorname{gap}_{\mathrm{ITA}} &\leq \frac{1}{2}\log\left(\frac{e}{2}\right) + \log\left(1 + (N-1)e^{-\frac{\operatorname{snrd}^2_{\min(X_D)}}{4}}\right) \quad \mathbf{DTD-ITA^{14-A}} \end{aligned}$$

Dytso, A.; Tuninetti, D.; Devroye, N., "On discrete alphabets for the twouser Gaussian interference channel with one receiver lacking knowledge of the interfering codebook," ITA, 2014, vol., no., pp.1,8, 9-14 Feb. 2014

Comparison of bounds

Input: PAM with number of points $N = \lfloor \sqrt{1 + \operatorname{snr}} \rfloor \Rightarrow H(X) = \log(N) \approx \frac{1}{2} \log(1 + \operatorname{snr})$



More recent tighter bounds

Tighter Ozarow-Wyner bounds on gaps available, based on the MMPE:

A. Dytso, R. Bustin, D. Tuninetti, N. Devroye, H.V. Poor and S. Shamai "On the Minimum Mean p-th Error in Gaussian channels and its Application," under submission to TransIT, 2016.

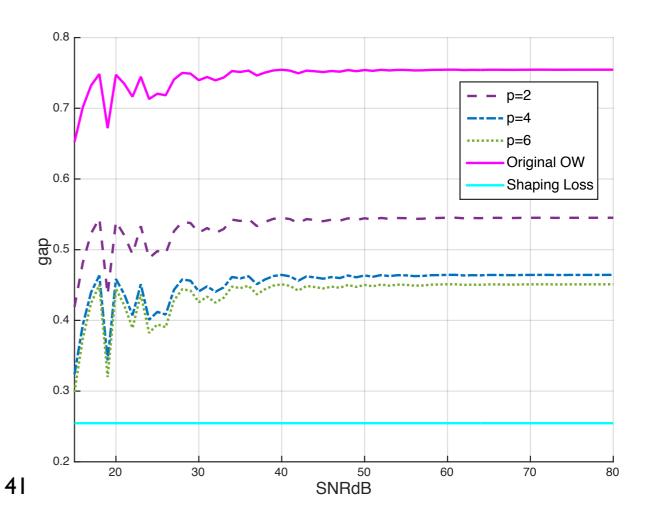
$$gap \leq \frac{1}{2} \log \left(\frac{\pi e}{6}\right) + \frac{1}{2} \log \left(1 + \frac{\operatorname{lmmse}(X, \operatorname{snr})}{d_{\min}(X_D)^2}\right),$$
$$\bigvee \operatorname{tighter}_{gap \leq \frac{1}{2} \log \left(\frac{\pi e}{6}\right) + \frac{1}{2} \log \left(1 + \frac{\operatorname{mmse}(X, \operatorname{snr})}{d_{\min}(X_D)^2}\right),$$
$$\bigvee \operatorname{tighter}_{distance}_{di$$

$$n^{-1}\mathsf{gap}_{\mathsf{p}} \leq \inf_{\mathbf{U}\in\mathcal{K}_{\mathsf{p}}} \left(G_{1,\mathsf{p}}(\mathbf{U},\mathbf{X}_{D}) + G_{2,\mathsf{p}}(\mathbf{U}) \right),$$

$$G_{1,\mathsf{p}}(\mathbf{U},\mathbf{X}_{D}) = \log \left(\frac{\|\mathbf{U}+\mathbf{X}_{D} - f_{\mathsf{p}}(\mathbf{X}_{D}|\mathbf{Y})\|_{\mathsf{p}}}{\|\mathbf{U}\|_{\mathsf{p}}} \right)$$

$$\int_{1}^{\text{for }\mathsf{p} \geq 1} \log \left(1 + \frac{\mathrm{mmpe}^{\frac{1}{\mathsf{p}}}(\mathbf{X}_{D},\mathsf{snr},\mathsf{p})}{\|\mathbf{U}\|_{\mathsf{p}}} \right),$$

$$G_{2,\mathsf{p}}(\mathbf{U}) = \log \left(\frac{k_{n,\mathsf{p}} \cdot n^{\frac{1}{\mathsf{p}}} \cdot \|\mathbf{U}\|_{\mathsf{p}}}{e^{\frac{1}{n}h_{\mathsf{e}}(\mathbf{U})}} \right).$$

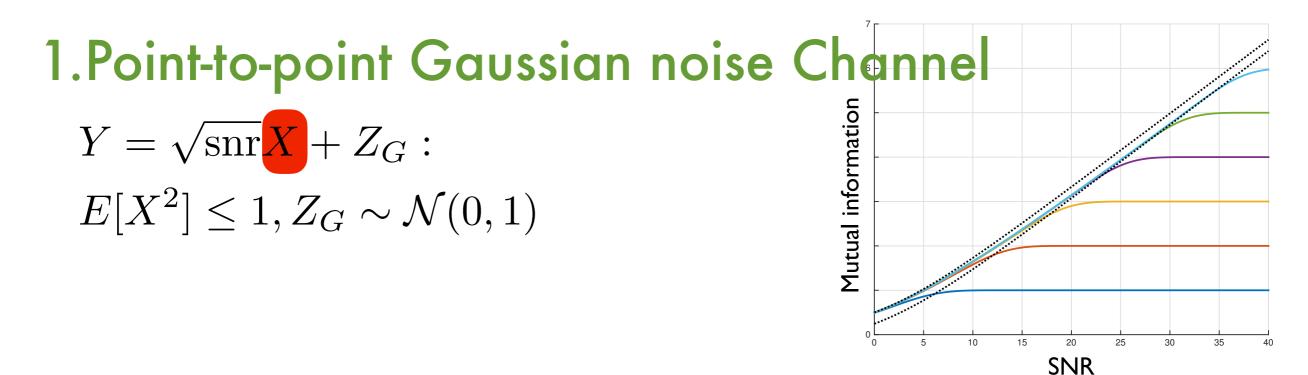


Why is discrete good?

good input

good interferer

Discrete is a good input.

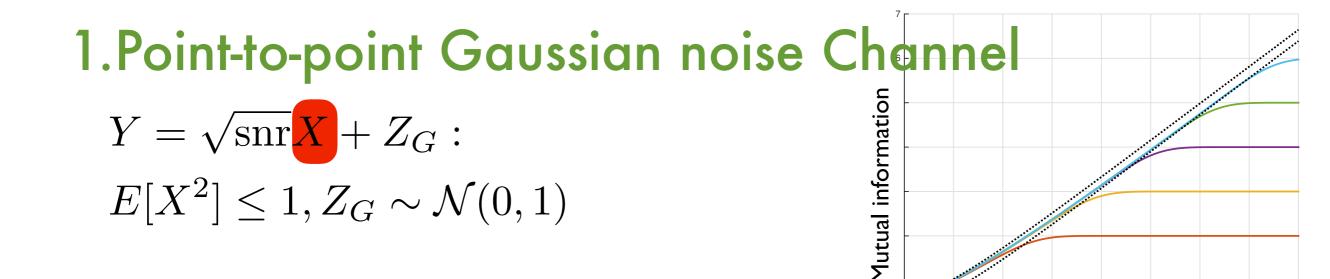


Capacity

$$C = \frac{1}{2}\log(1 + \operatorname{snr})$$

achieved by Gaussian

Discrete is a good input.



Capacity

$$C = \frac{1}{2}\log(1 + \operatorname{snr})$$

achieved by Gaussian

with PAM:

$$N = \lfloor \sqrt{1 + \operatorname{snr}} \rfloor$$
$$C \ge \frac{1}{2} \log(1 + \operatorname{snr}) - \operatorname{gap}$$
$$\operatorname{gap} = \frac{1}{2} \log\left(\frac{4\pi e}{3}\right)$$

SNR

Discrete is a good interferer.

2.Point-to-point Gaussian noise Channel with State $Y = \sqrt{\operatorname{snr}}X + hT + Z_G$: $E[X^2] \le 1, Z_G \sim \mathcal{N}(0, 1),$ $T \sim \operatorname{discrete:} |T| = N \text{ and } d^2_{\min(T)} > 0$

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Discrete Interference

$$C \ge I(X_G; \sqrt{\operatorname{snr}} X_G + hT + Z_G)$$

$$\ge \frac{1}{2} \log(1 + \operatorname{snr}) - \operatorname{gap}$$

$$\operatorname{gap} = \frac{1}{2} \log\left[\frac{2\pi e}{12} \left(1 + \frac{12}{d_{\min}^2(T)} \frac{|h|^2 \mathcal{E}_T}{|h|^2 \mathcal{E}_T + 1 + \operatorname{snr}}\right)\right]$$

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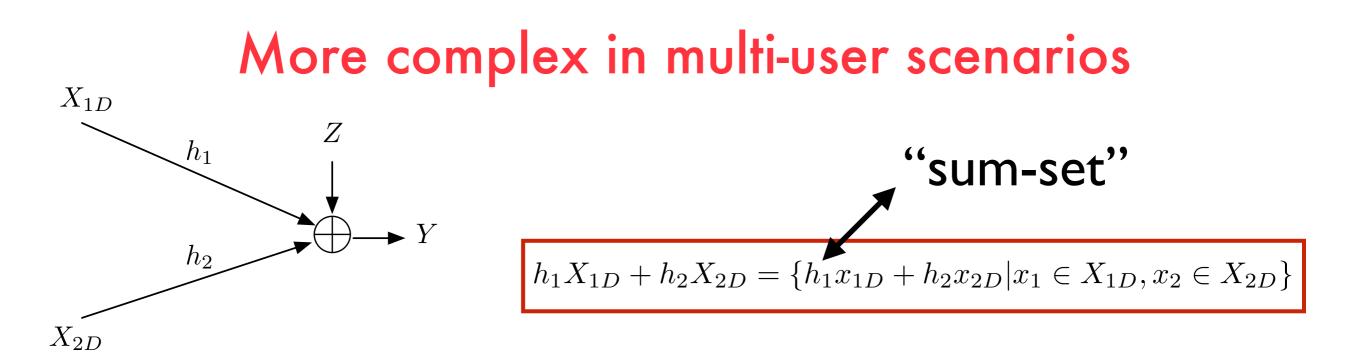
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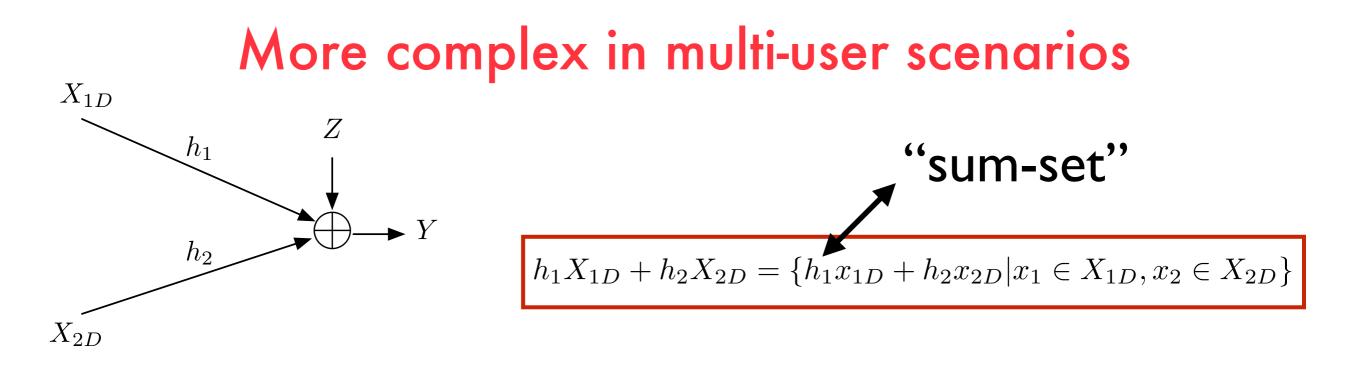
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$$= \frac{1}{2} \log \left(1 + \frac{\operatorname{snr}}{1 + |h|^2 \mathcal{E}_T} \right)$$

Gaussian Interference

Discrete inputs in multi-user channels



Discrete inputs in multi-user channels



$$|h_1 X_{1D} + h_2 X_{2D}| = |\{h_1 x_{1D} + h_2 x_{2D} | x_1 \in X_{1D}, x_2 \in X_{2D}\}|$$
???
$$d_{\min}(h_1 X_{1D} + h_2 X_{2D}) = \min\{|s_i - s_j| : s_i, s_j \in h_1 X_{1D} + h_2 X_{2D}, i \neq j\}$$
???

New phenomenon

Example, BPSK:

$$X_{1D} = X_{2D} = \{-1, +1\}$$

$$h_1 X_{1D} + h_2 X_{2D} \stackrel{(h_1 = 1, h_2 = 2)}{=} \{3, -1, 1, 3\}$$
$$\stackrel{(h_1 = 1, h_2 = 1)}{=} \{1, 0, -1\}$$

"Cardinality is Sensitive to Channel Gain Values."

Overall proposition / tool

• cardinality of the sum-set $\{h_x X + h_y Y\}$

Proposition: Let $X \sim \mathsf{PAM}(|X|, d_{\min(X)})$ and $Y \sim \mathsf{PAM}(|Y|, d_{\min(Y)})$. Then for $(h_x, h_y) \in \mathbb{R}^2$

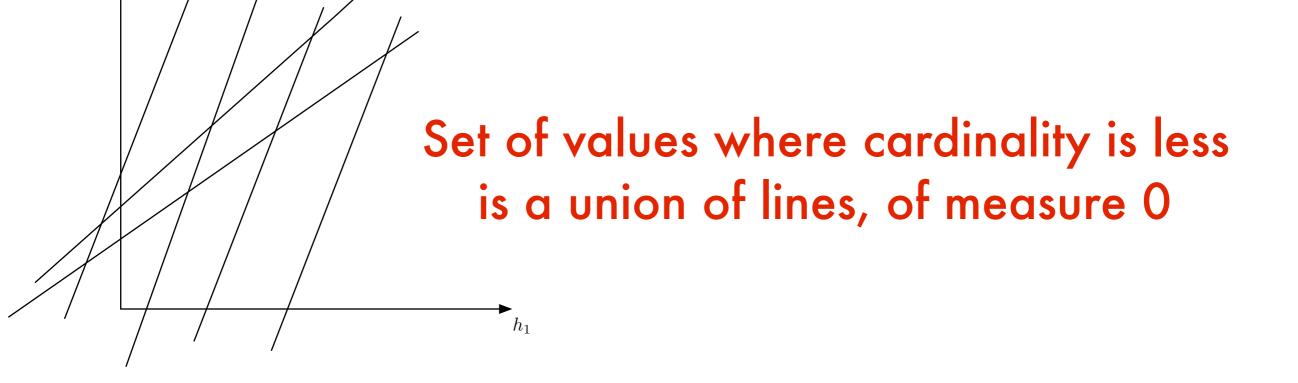
 $|h_x X + h_y Y| = |X||Y| \text{ almost everywhere (a.e.)}, \tag{1}$

• minimum distance of the sum-set

and
$$d_{\min(h_x X + h_y Y)} \ge \dots$$
?

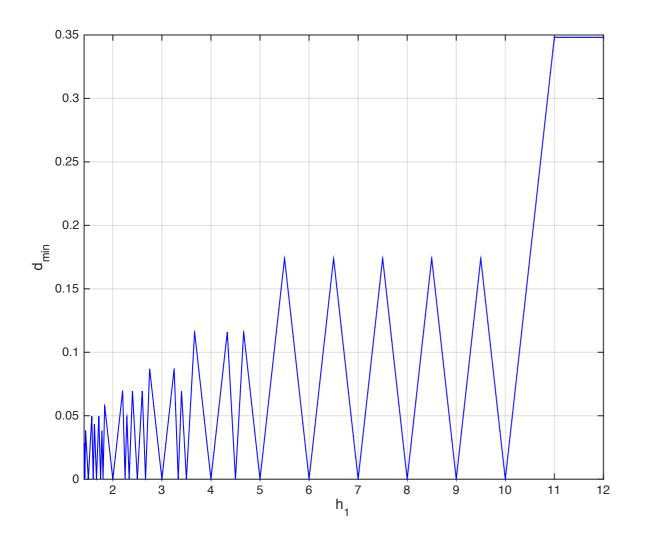
Cardinality

 $|h_x X + h_y Y| = |X||Y|$ almost everywhere (a.e.)



Minimum distance

Example:h2=1, N1=N2=10



Very Irregular

Can we even have a lower bound?

$$\operatorname{gap}_{\text{OW-B}} \le \frac{1}{2} \log\left(\frac{\pi e}{6}\right) + \frac{1}{2} \log\left(1 + \frac{12}{\operatorname{snr} d_{\min(X_D)}^2}\right)$$

Minimum distance, case 1: no overlap

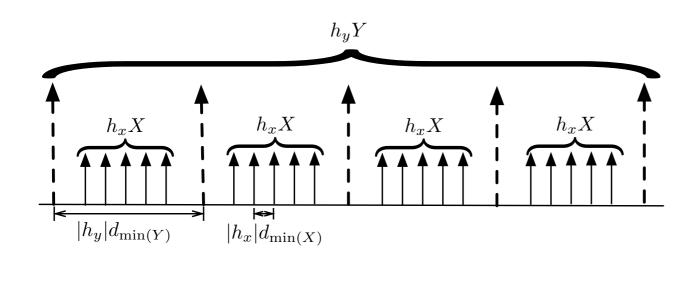
We have

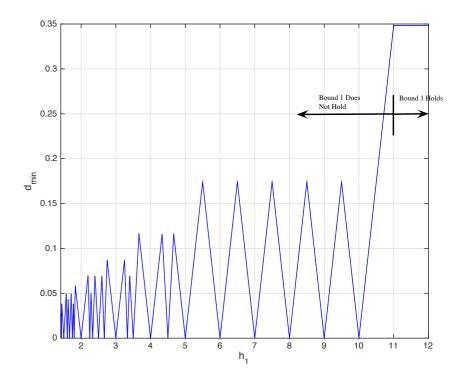
$$d_{\min(h_x X + h_y Y)} = \min(|h_x|d_{\min(X)}, |h_y|d_{\min(Y)})$$

under the following conditions

either
$$|Y||h_y|d_{\min(Y)} \le |h_x|d_{\min(X)}$$
,
or $|X||h_x|d_{\min(X)} \le |h_y|d_{\min(Y)}$ (shown below).

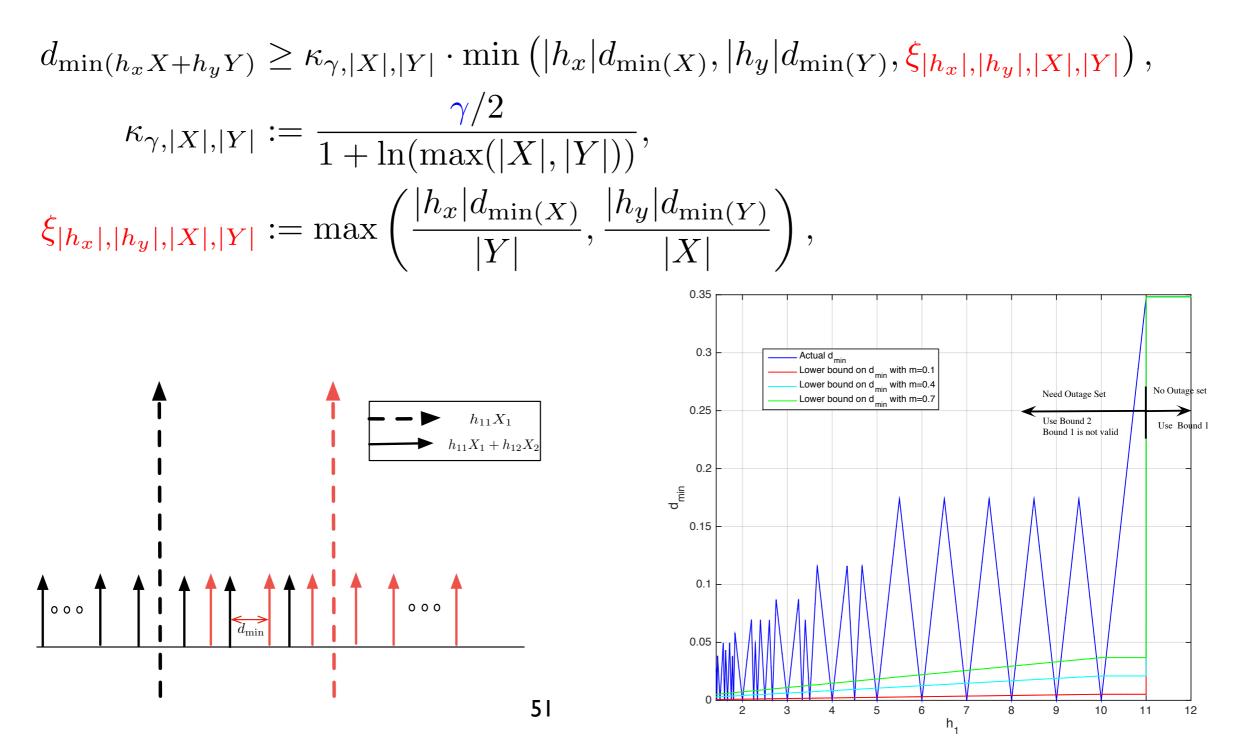
50



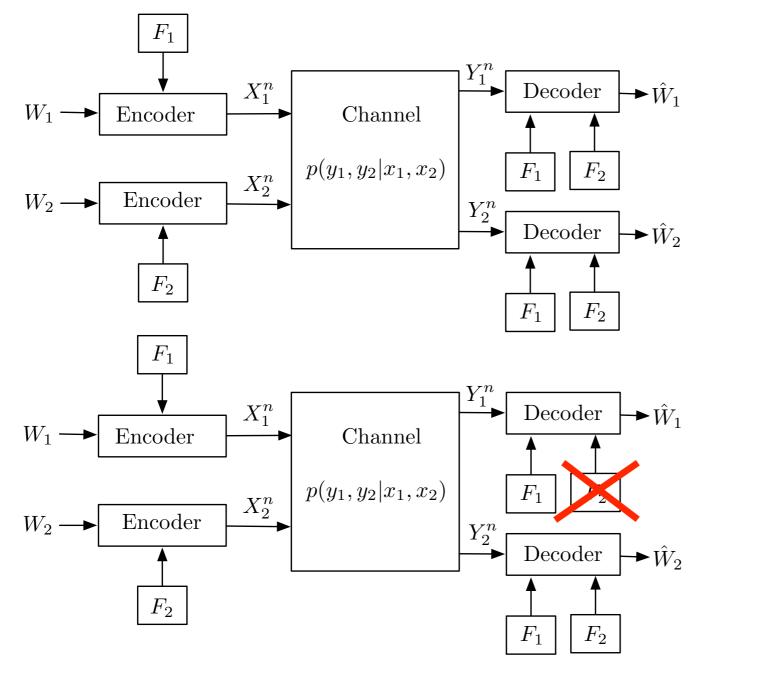


Minimum distance, case 2: with overlap

Then, up to a set of (h_x, h_y) of measure no more than γ , we have



Applications of discrete inputs

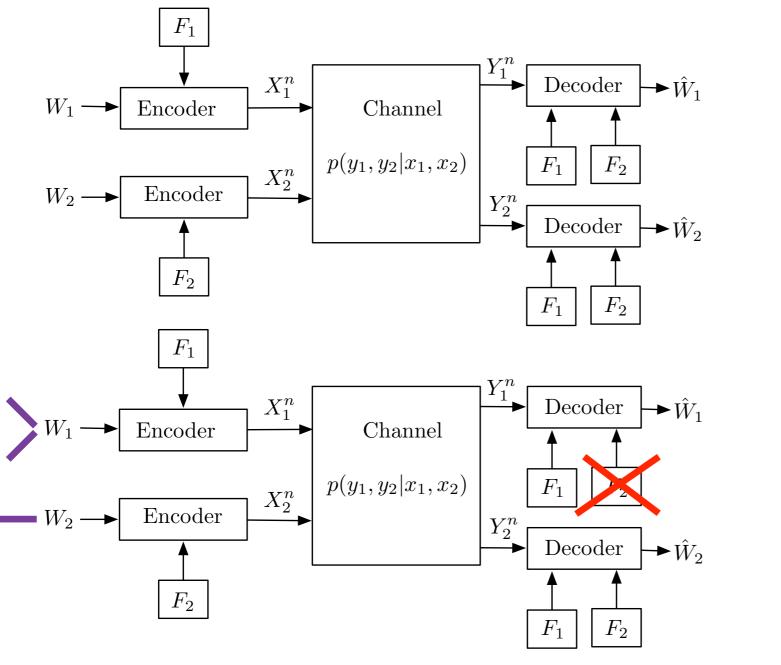


HK+Gaussian Inputs 1/2 bit

R. Etkin, D. Tse, and H. Wang, "Gaussian interference channel capacity to within one bit," IEEE Trans. Inf. Theory, vol. 54, no. 12, pp. 5534–5562, Dec. 2008.

"One-sided" HK+ Mixed Inputs 3.34 bits

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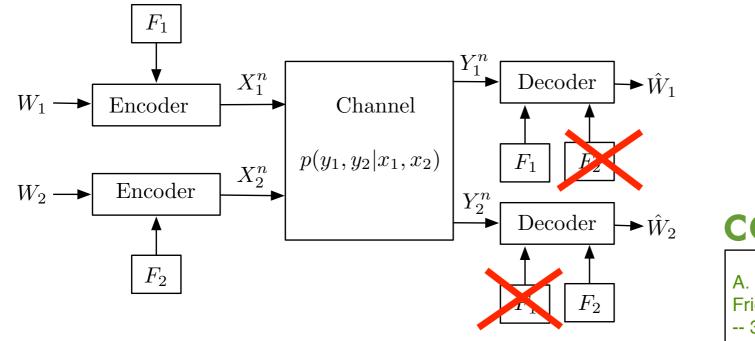


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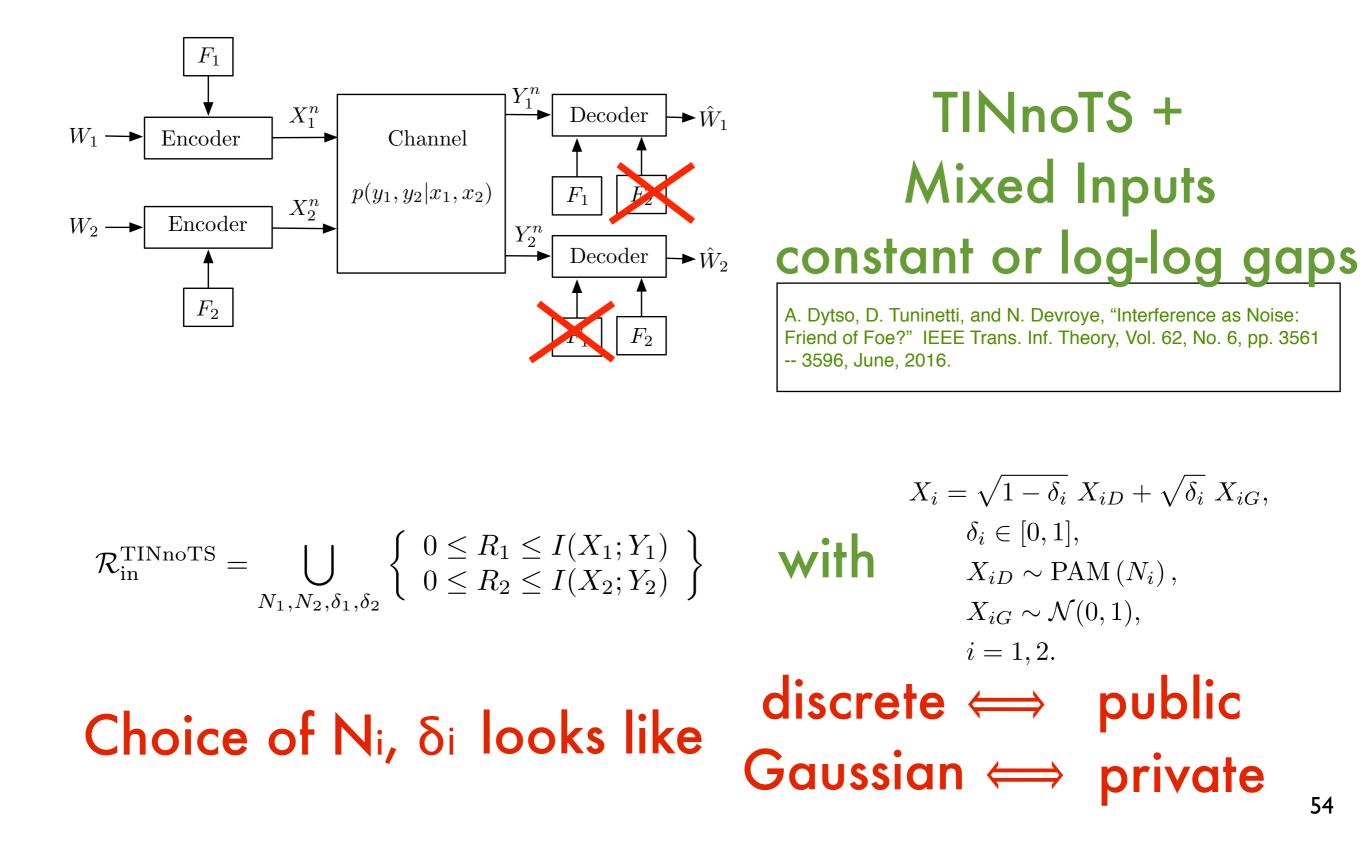
TINnoTS + Mixed Inputs Constant or log-log gaps A. Dytso, D. Tuninetti, and N. Devroye, "Interference as Noise:

Friend of Foe?" IEEE Trans. Inf. Theory, Vol. 62, No. 6, pp. 3561 -- 3596, June, 2016.

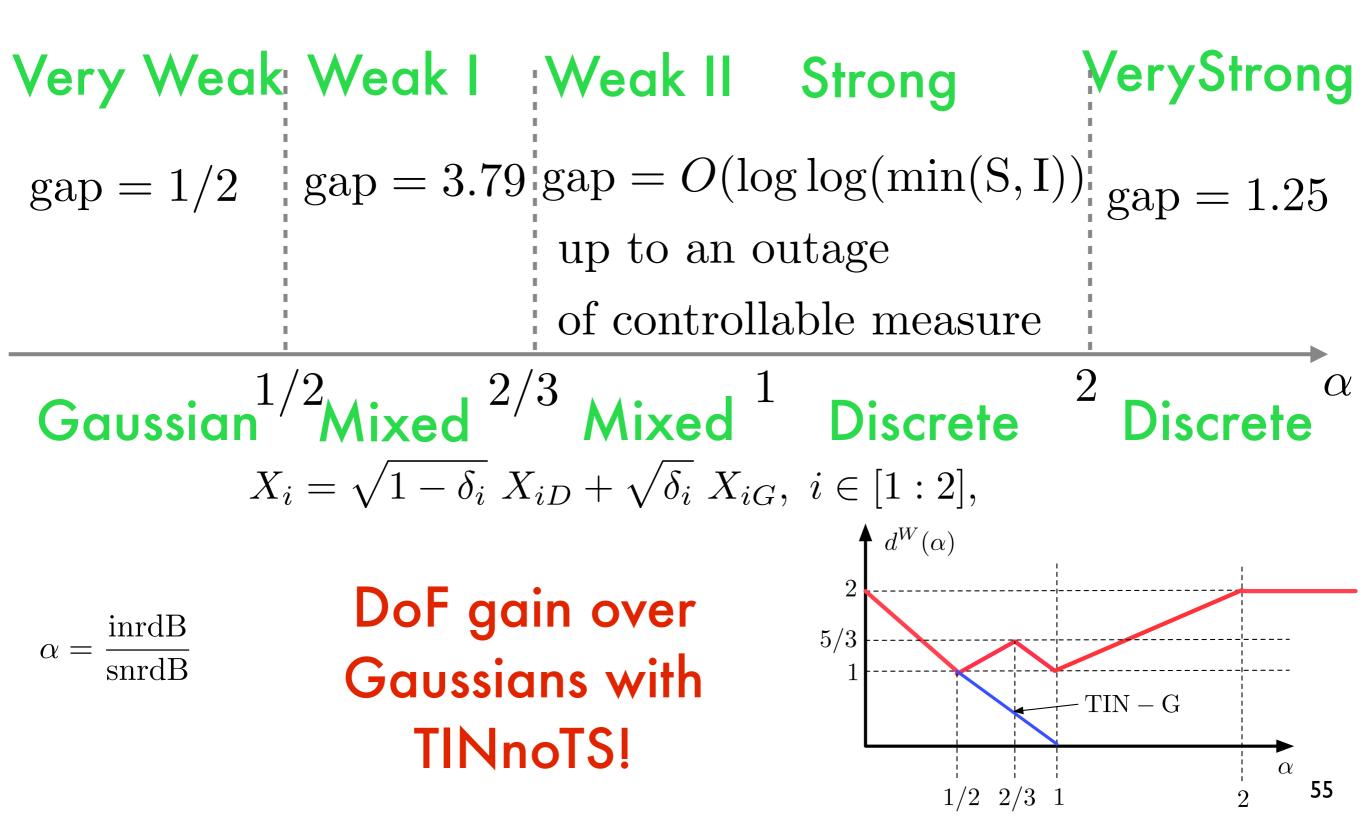
$$\mathcal{R}_{\text{in}}^{\text{TINnoTS}} = \bigcup_{N_1, N_2, \delta_1, \delta_2} \left\{ \begin{array}{l} 0 \le R_1 \le I(X_1; Y_1) \\ 0 \le R_2 \le I(X_2; Y_2) \end{array} \right\}$$

 $X_{i} = \sqrt{1 - \delta_{i}} X_{iD} + \sqrt{\delta_{i}} X_{iG},$ $\delta_{i} \in [0, 1],$ $X_{iD} \sim \text{PAM}(N_{i}),$ $X_{iG} \sim \mathcal{N}(0, 1),$ i = 1, 2.

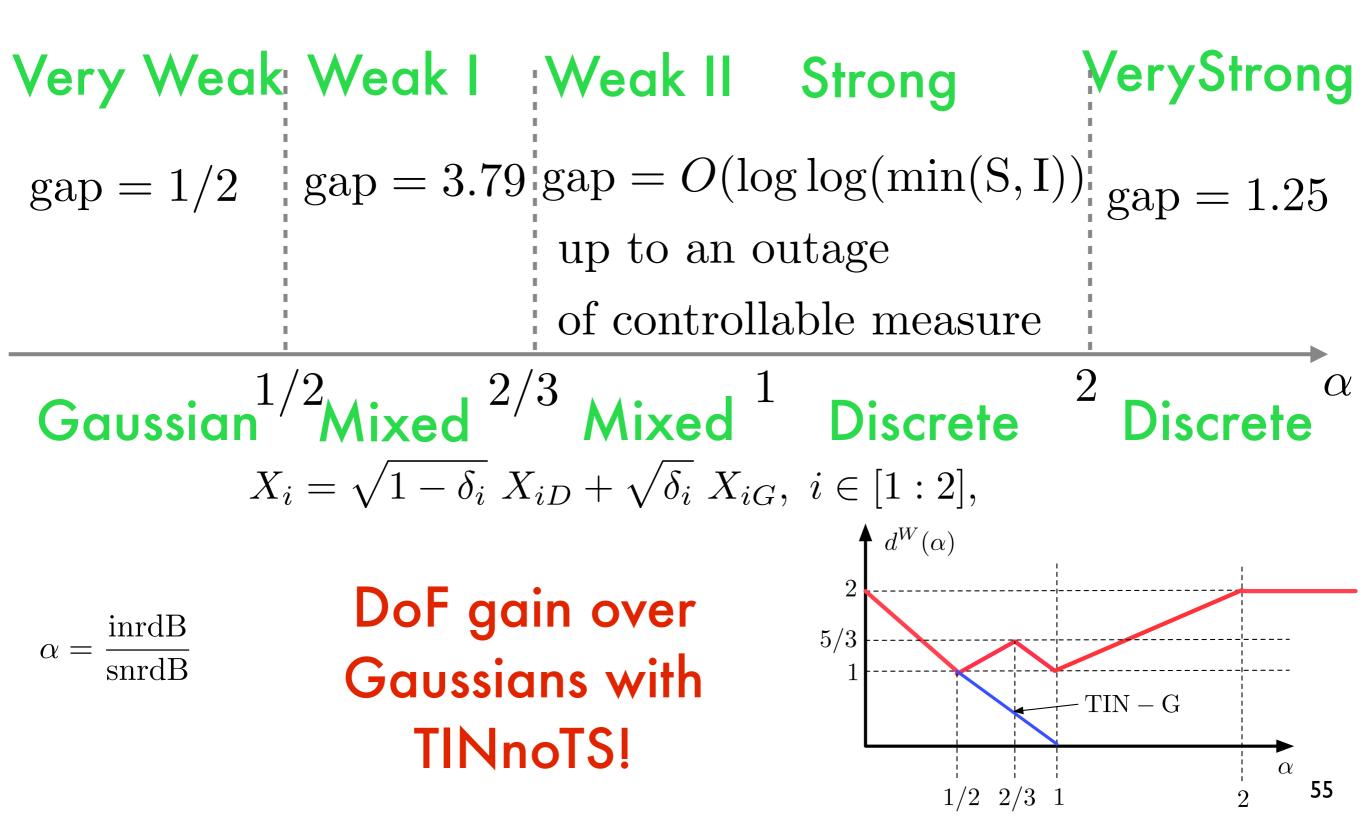
with



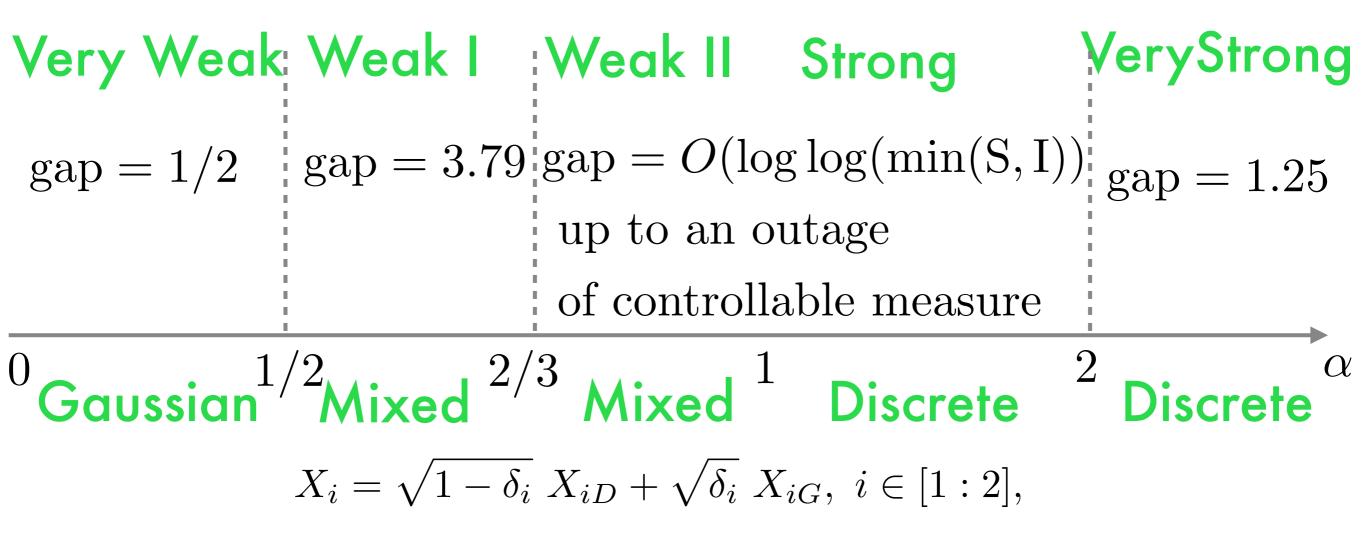
Approximate optimality of TINnoTS in Gaussian-IC



Approximate optimality of TINnoTS in Gaussian-IC



Approximate optimality of TINnoTS in Gaussian-IC



 $\alpha = \frac{\text{inrdB}}{\text{snrdB}}$

<u>Closed-form</u> expressions for number of points, power splits and gap 56

- use non-Gaussian inputs: good inputs, good interferers
- general tools on bounding dmin, mutual information applicable elsewhere?
- mixed inputs hence approximately optimal for the codebook oblivious G-IC

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- **OPEN:** can we develop a smart set of multi-letter discrete inputs and evaluate these in the capacity achieving expression for the G-IC?

Capacity:
$$C = \lim_{n \to \infty} co\left(\bigcup_{P_{X_1^n X_2^n} = P_{X_1^n} P_{X_2^n}} \left\{ \begin{array}{l} 0 \le R_1 \le \frac{1}{n} I(X_1^n; Y_1^n) \\ 0 \le R_2 \le \frac{1}{n} I(X_2^n; Y_2^n) \end{array} \right\} \right)$$

R. Ahlswede, "Multi-way communication channels," in Proc. IEEE Int. Symp. Inf. Theory, March 1973, pp. 23–52.

Discussion: relation to.....

Characterizing when (low INR) TIN is optimal in larger interference networks:

[Chunhua Geng, Syed A. Jafar, On the Optimality of Treating Interference as Noise: Compound Interference Networks, IEEE Trans. on Info. Theory, Accepted, 2016.]

[Hua Sun, Syed A. Jafar, On the Optimality of Treating Interference as Noise for K user Parallel Gaussian Interference Networks , IEEE Transactions on Information Theory, Vol. 62, No. 4, Pages: 1911-1930, April 2016.]

[Chunhua Geng, Syed A. Jafar, On the Optimality of Treating Interference as Noise: General Message Sets, IEEE Transactions on Information Theory, Vol. 61, No. 7, Pages: 3722-3736, July 2015.]

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Using point-to-point codes in interference networks:

[F. Baccelli, A. El Gamal, D. Tse, "Interference Networks with Point-to-Point Codes," IEEE Trans. on Info. Theory, Vol. 57, No. 5, May 2011.]

[]. Sebastian, C. Karakus, S. Diggavi "Approximately achieving the feedback interference channel capacity with point-to-point codes" ISIT 2016.]

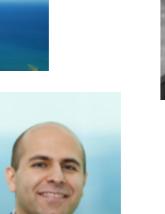
[Young Han Kim et al. http://circuit.ucsd.edu/~yhk/pdfs/swcm.pdf]

[B. Bandemer, A. El Gamal, Y.-H. Kim, "Optimal achievable rates for interference networks with random codes" Trans IT 2015.]















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US Our message: use discrete inputs and TIN in interference networks for FBA simple comm. where do not have synchronism or codebook knowledge Point week TENE

[]. Sebastian, C. Karakus, S. Diggavi "Approximately achieving the feedback interference channel capacity with point-to-point codes" ISIT 2016.]

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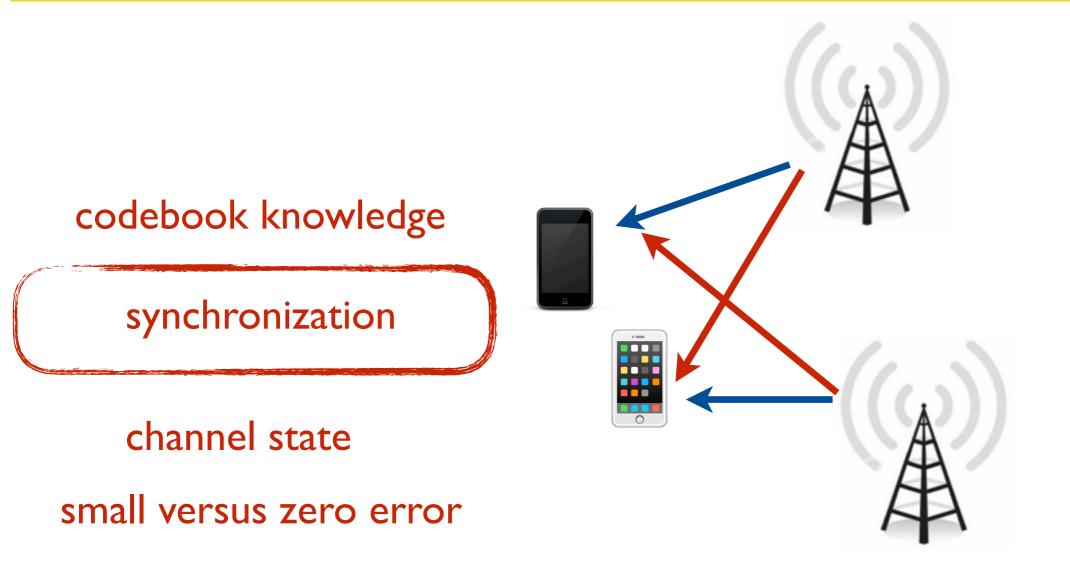
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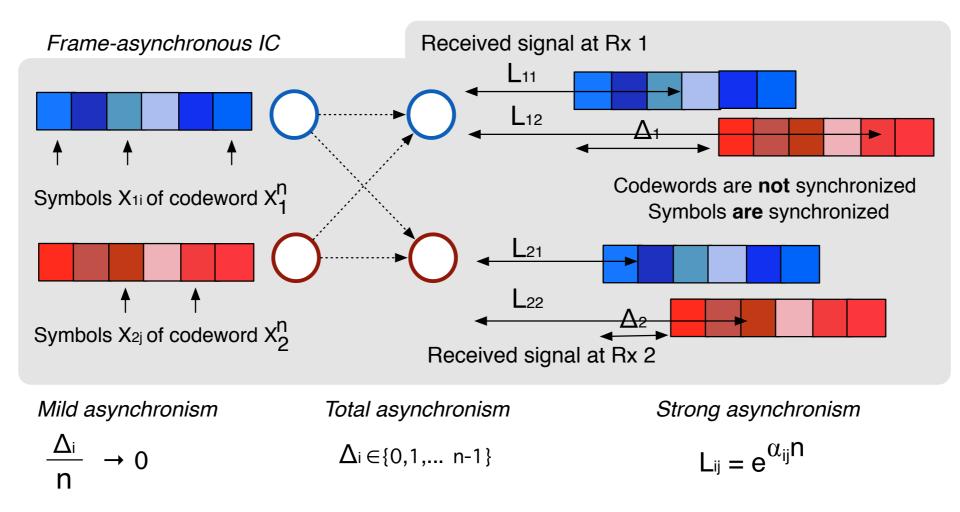






Approximate capacity of ICs with lack of synchronization

• in networks, often assume all nodes are synchronized



•this may be unrealistic sometimes....

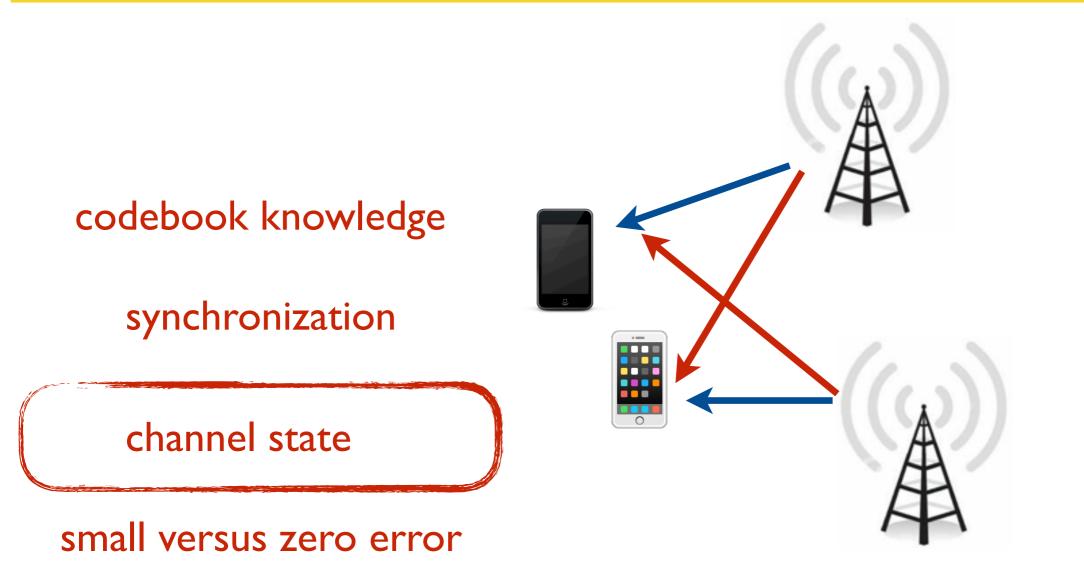
Approximate capacity of ICs with total asynchronism

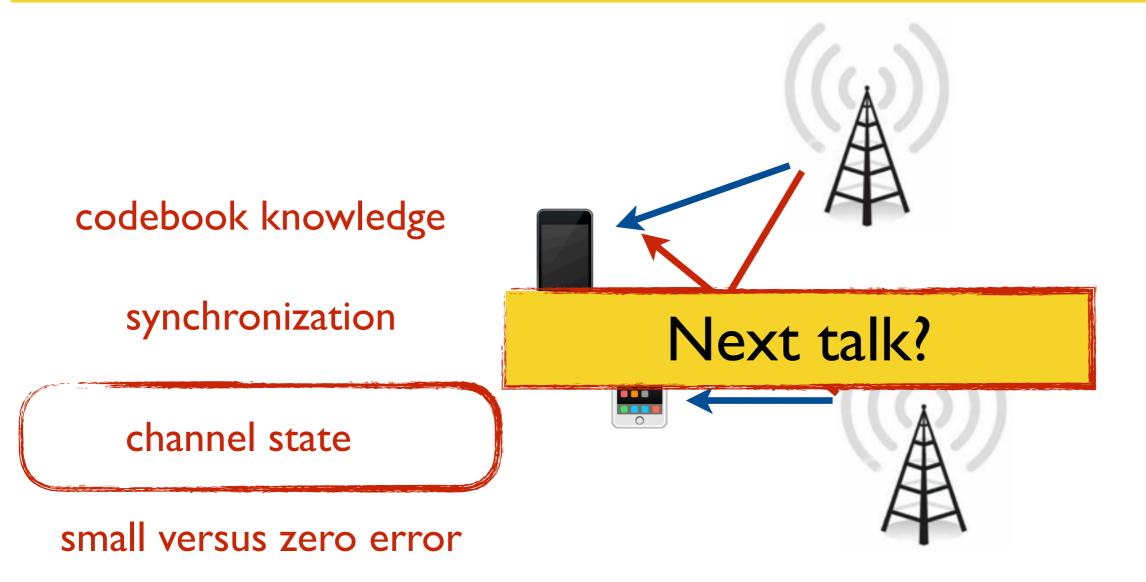
Treat Interference as Noise without Time Sharing Inner Bound:

$$\mathcal{R}_{\text{in}}^{\text{TINnoTS}} = \bigcup_{P_{X_1 X_2} = P_{X_1} P_{X_2}} \left\{ \begin{array}{l} 0 \le R_1 \le I(X_1; Y_1) \\ 0 \le R_2 \le I(X_2; Y_2) \end{array} \right\}$$

[E. Calvo, J.R. Fonollosa, J.Vidal, ``On the Totally Asynchronous Interference Channel with Single-User Receivers'' ISIT 2009.]

 this is achievable by asynchronous G-IC, so our approximate gap to capacity results apply even without synchronization!







Network coding: [T. Chan and A. Grant, ``On capacity regions of non-multicast networks" ISIT 2010]

[M. Langberg, M. Effros, ``Network coding: is zero error always possible?" Allerton 2011.]

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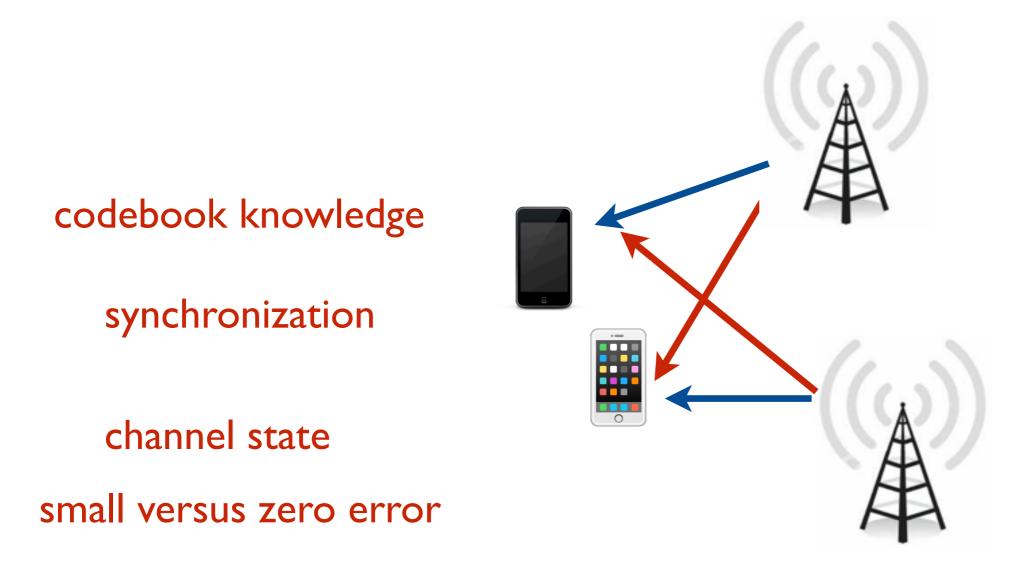
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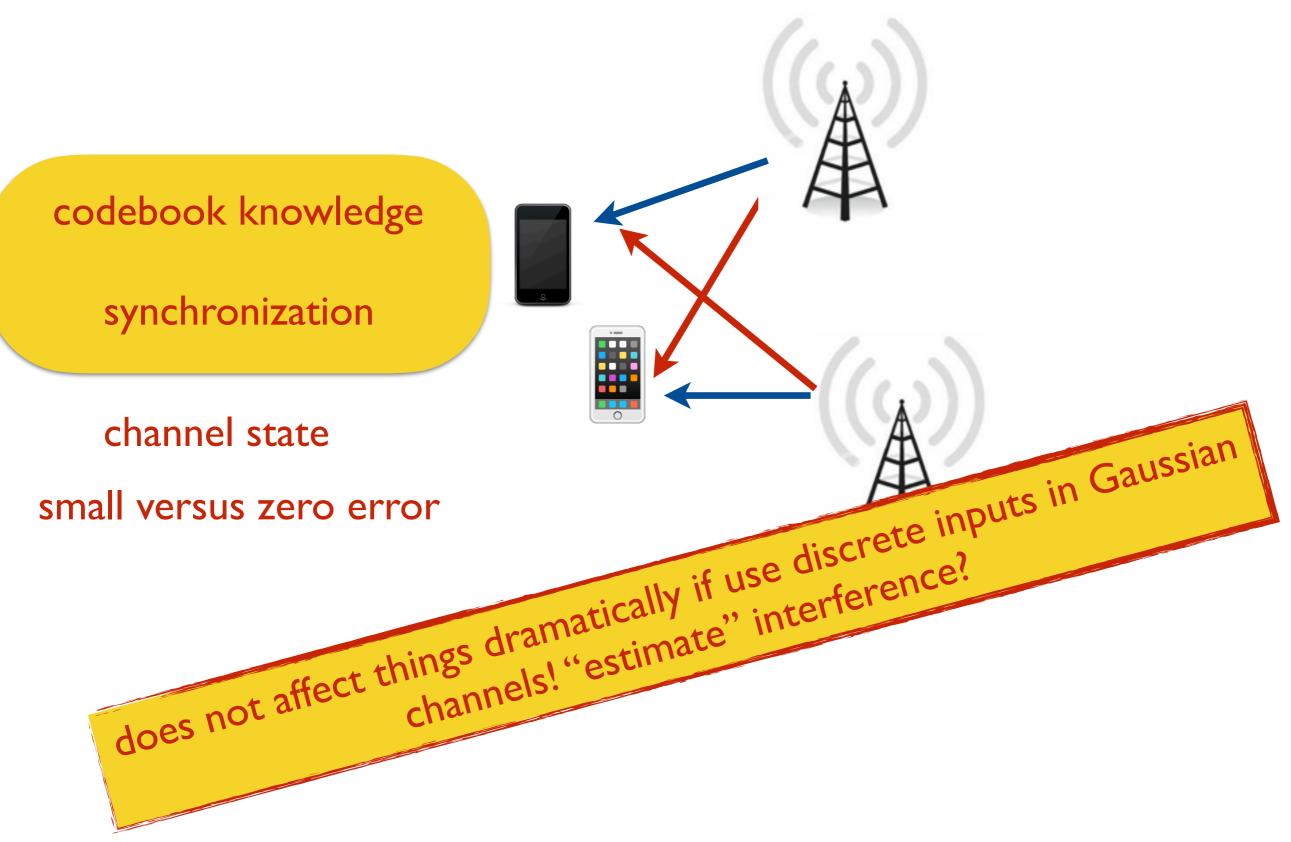
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[Y. Chen, N. Devroye "On the optimality of Color-and-Forward Relaying for a Class of Zero-error Primitive Relay Channels?" ISIT 2015, under submission to TransIT]

CONCLUSION FOR INTERFERENCE CHANNEL

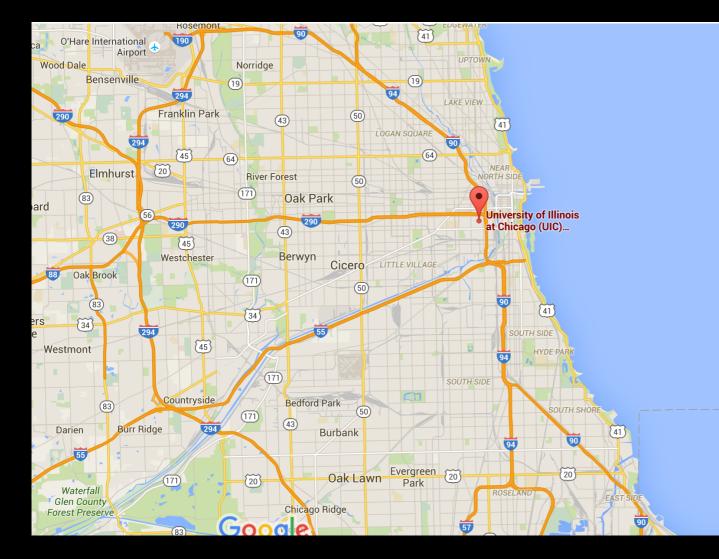


CONCLUSION FOR INTERFERENCE CHANNEL





UIC is a public school in downtown Chicago





UIC is a great place to visit

- home of the "NICEST" lab

UIC



Networks Information Communications

Natasha Devroye

67

Hulya Seferoglu Besma I Smida Tu

Daniela Tuninetti

- home of the best "Brutalist"
 architecture in the world









Questions + discussions now, later, email are always welcome

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